Digital Design & Computer Arch.

Lecture 4: Combinational Logic I

Prof. Onur Mutlu

ETH Zürich

Spring 2021

5 March 2021

Assignment: Required Readings

- This week
  - Combinational Logic
    - P&P Chapter 3 until 3.3 + H&H Chapter 2

- Next week
  - Hardware Description Languages and Verilog
    - H&H Chapter 4 until 4.3 and 4.5
  - Sequential Logic
    - P&P Chapter 3.4 until end + H&H Chapter 3 in full

- By the end of next week, make sure you are done with
  - **P&P Chapters 1-3 + H&H Chapters 1-4**

A Note on Hardware vs. Software

- This course might seem like it is only "Computer Hardware"
- However, you will be much more capable if you master both hardware and software (and the interface between them)
  - Can develop better software if you understand the hardware
  - Can design better hardware if you understand the software
  - Can design a better computing system if you understand both
- This course covers the HW/SW interface and microarchitecture
  - We will focus on tradeoffs and how they affect software
- Recall the four mysteries

Recap: Four Mysteries

- Meltdown & Spectre (2017-2018)
- Rowhammer (2012-2014)
- Memory Performance Attacks (2006-2007)
- Memories Forget: Refresh & RAIDR (2011-2012)

Computer Architecture as an Enabler of the Future

Assignment: Required Lecture Video

- Why study computer architecture? Why is it important?
- Future Computing Platforms: Challenges & Opportunities

- **Required Assignment**
  - **Watch one of** Prof. Mutlu's lectures and analyze either (or both)
  - https://www.youtube.com/watch?v=kgiZlSOcGFM (May 2017)
  - https://www.youtube.com/watch?v=mskTeNnf-i0 (Feb 2021)

- **Optional Assignment – for 1% extra credit**
  - **Write a 1-page summary** of one of the lectures and email us
    - What are your key takeaways?
    - What did you learn?
    - What did you like or dislike?
    - Submit your summary to Moodle

… but, first …

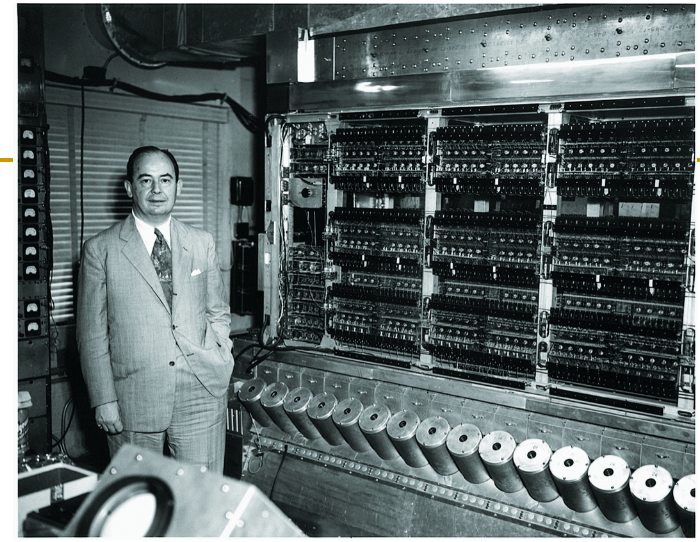
- Let’s understand the fundamentals…

- You can change the world only if you understand it well enough…
  - Especially the basics (fundamentals)
  - Past and present dominant paradigms
  - And, their advantages and shortcomings – tradeoffs
  - And, what remains fundamental across generations
  - And, what techniques you can use and develop to solve problems

Fundamental Concepts

What is A Computer?

- Three key components
- Computation
- Communication
- Storage/memory



Burks, Goldstein, von Neumann, "Preliminary discussion of the logical design of an electronic computing instrument," 1946.

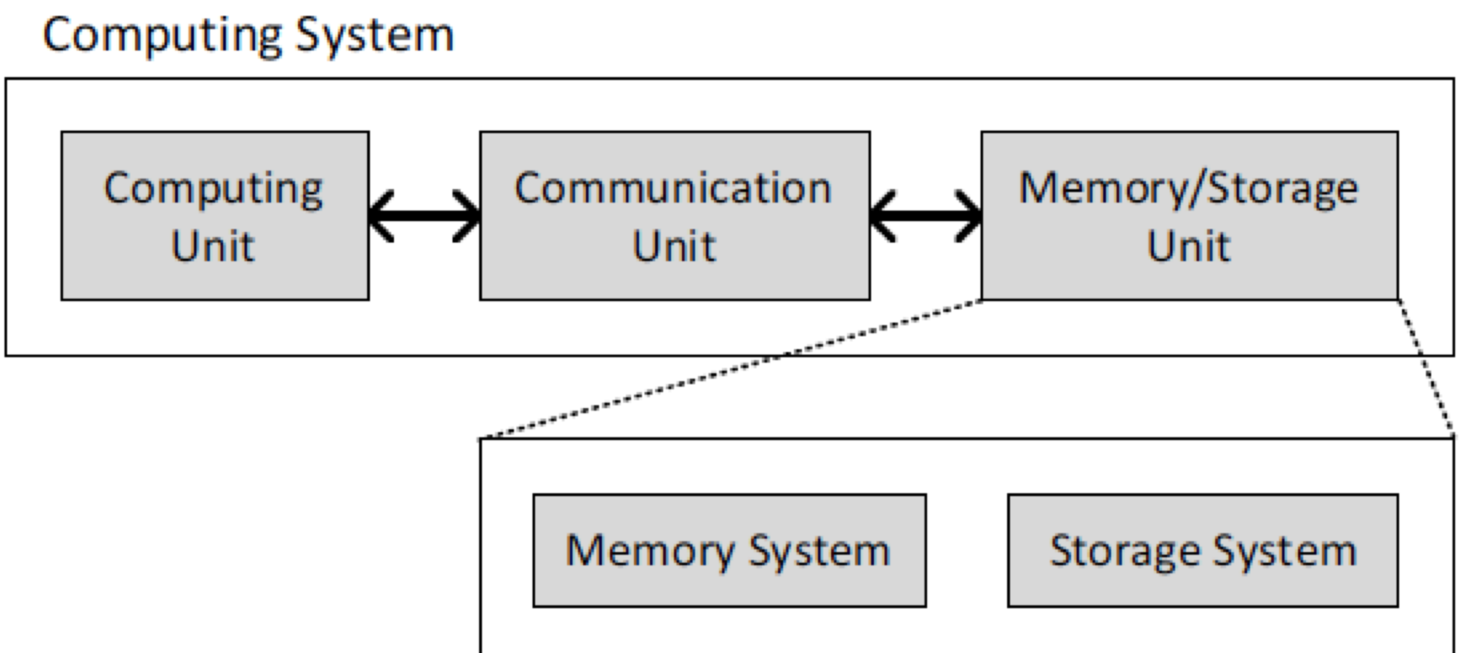
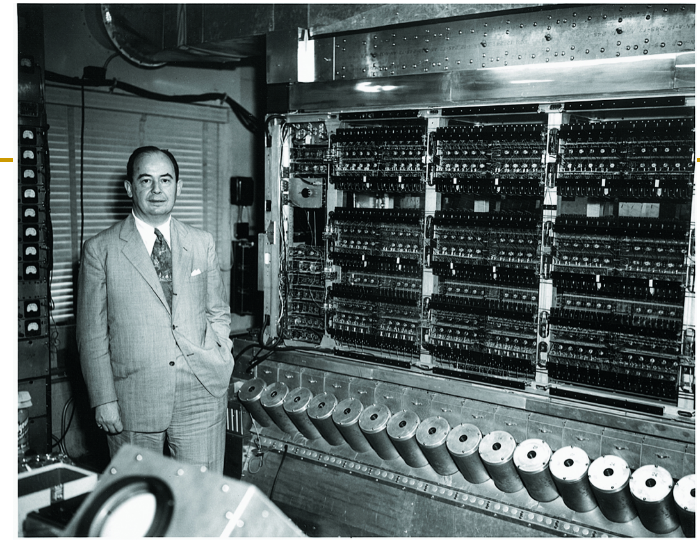


Image source: https://lbsitbytes2010.wordpress.com/2013/03/29/john-von-neumann-roll-no-15/

What is A Computer?

- Three key components
- Computation
- Communication
- Storage/memory



Burks, Goldstein, von Neumann, "Preliminary discussion of the logical design of an electronic computing instrument," 1946.

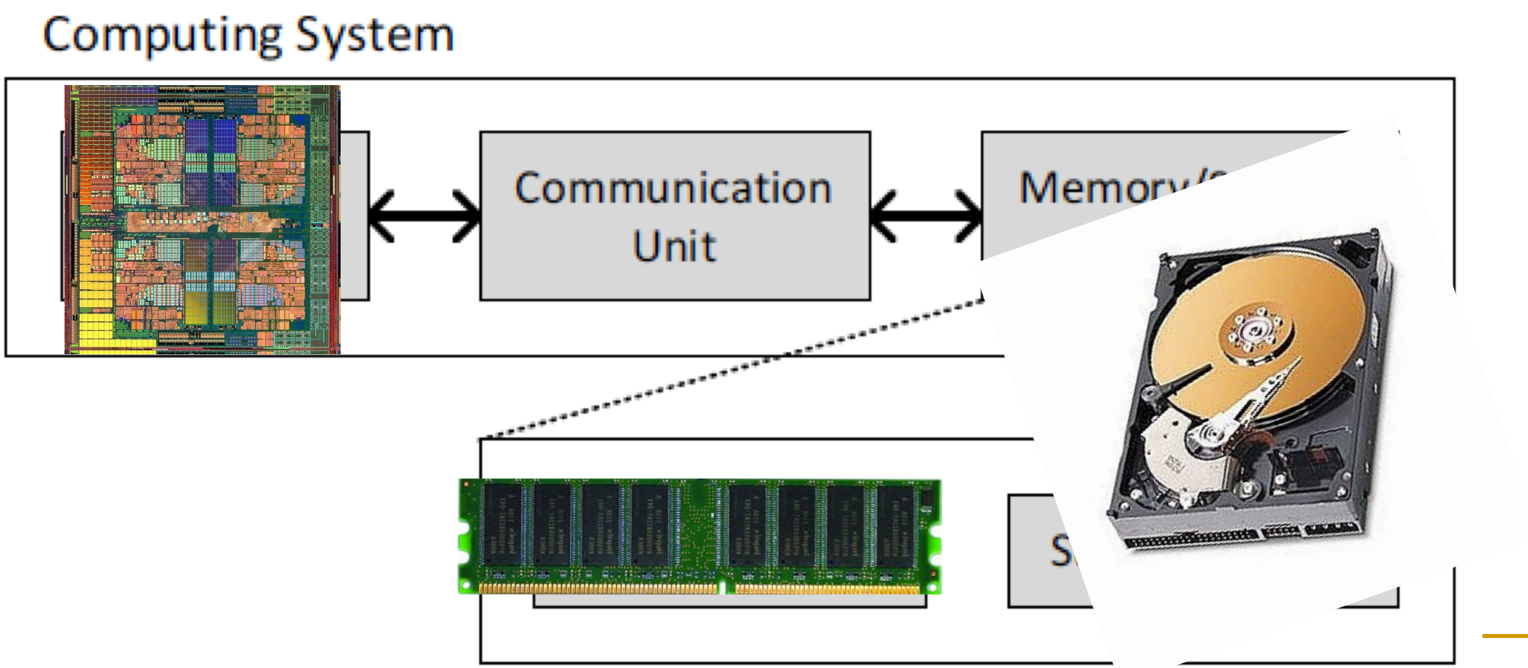
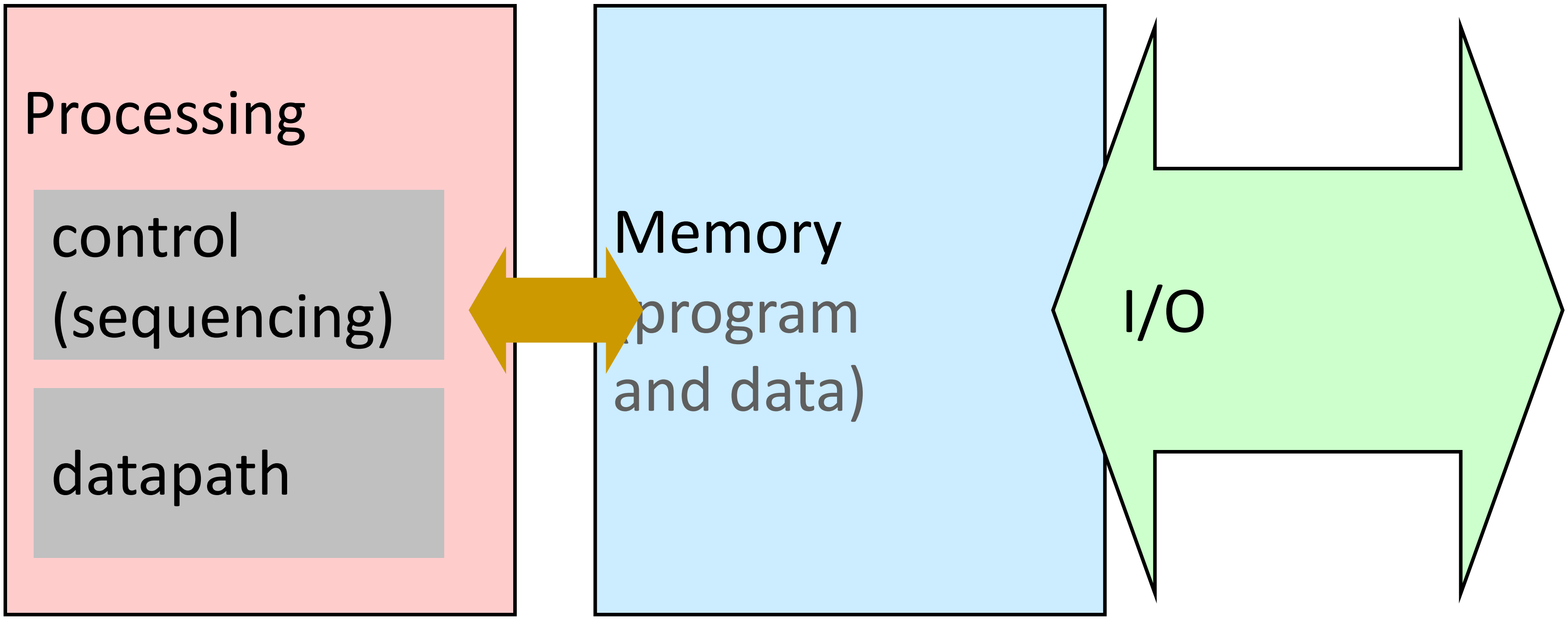


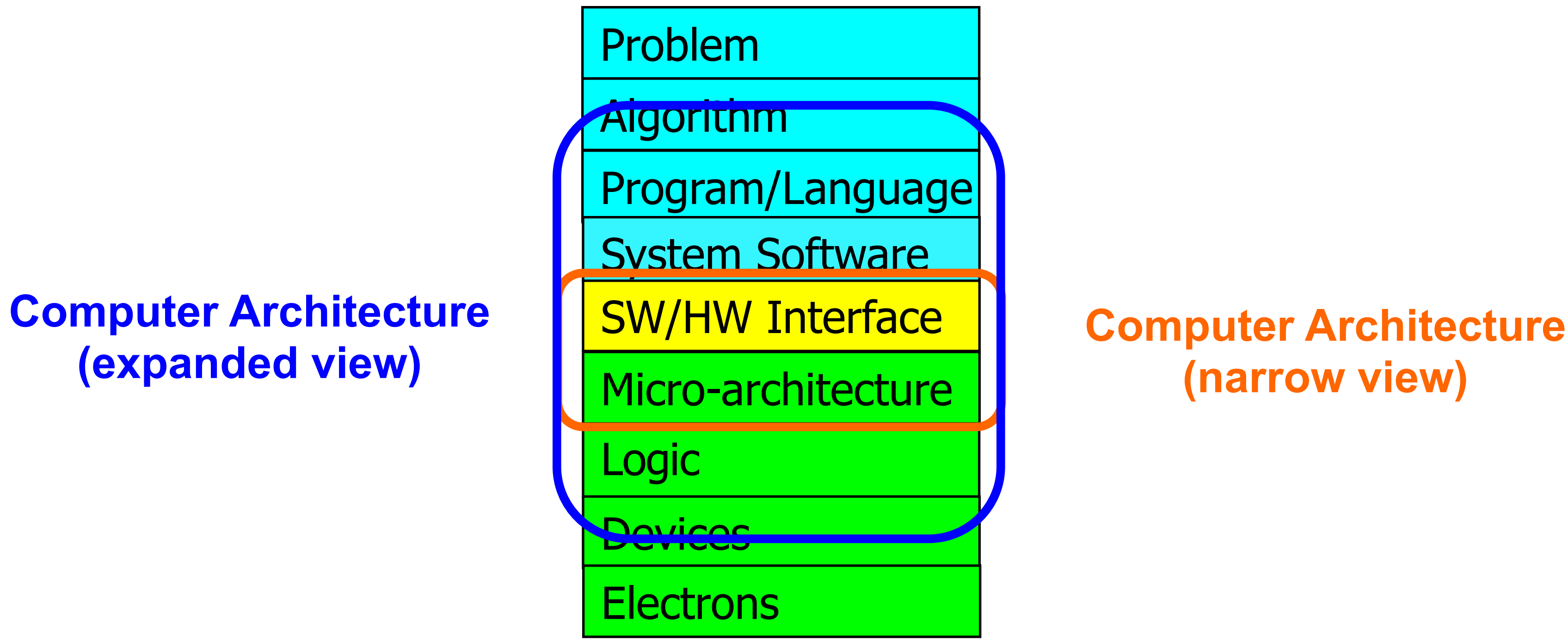
Image source: https://lbsitbytes2010.wordpress.com/2013/03/29/john-von-neumann-roll-no-15/

What is A Computer?

- We will cover all three components

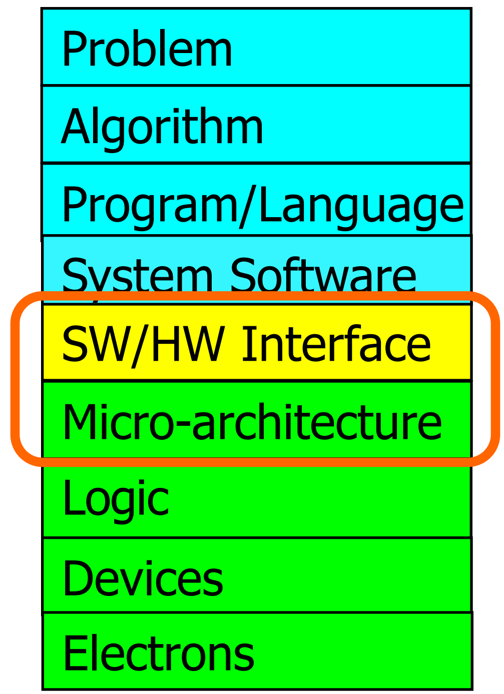


Recall: The Transformation Hierarchy



What We Will Cover (I)

- Combinational Logic Design
- Hardware Description Languages (Verilog)
- Sequential Logic Design
- Timing and Verification
- ISA (MIPS and LC3b)
- MIPS Assembly Programming

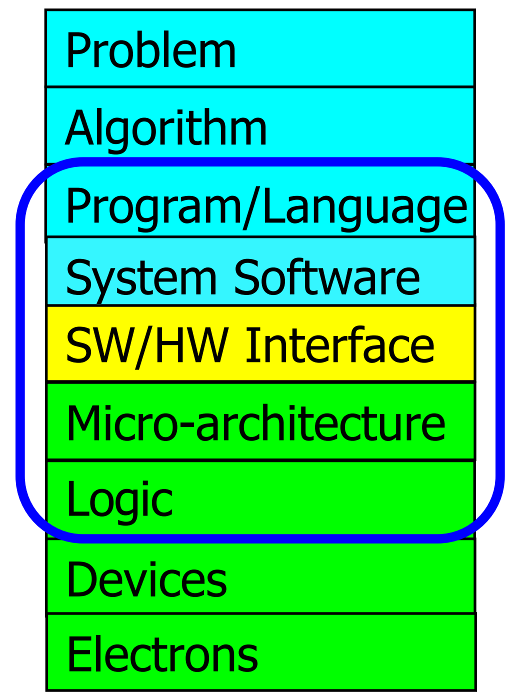


What We Will Cover (II)

- Microarchitecture Basics: Single-cycle
- Multi-cycle and Microprogrammed Microarchitectures
- Pipelining
- Issues in Pipelining: Control & Data Dependence Handling, State Maintenance and Recovery, …
- Out-of-Order Execution
- Other Processing Paradigms (SIMD, VLIW, Systolic, …)
- Memory and Caches
- Virtual Memory

Processing Paradigms We Will Cover

- Pipelining
- Out-of-order execution
- Dataflow (at the ISA level)
- Superscalar Execution
- VLIW
- SIMD Processing (Vector & Array, GPUs)
- Decoupled Access-Execute
- Systolic Arrays

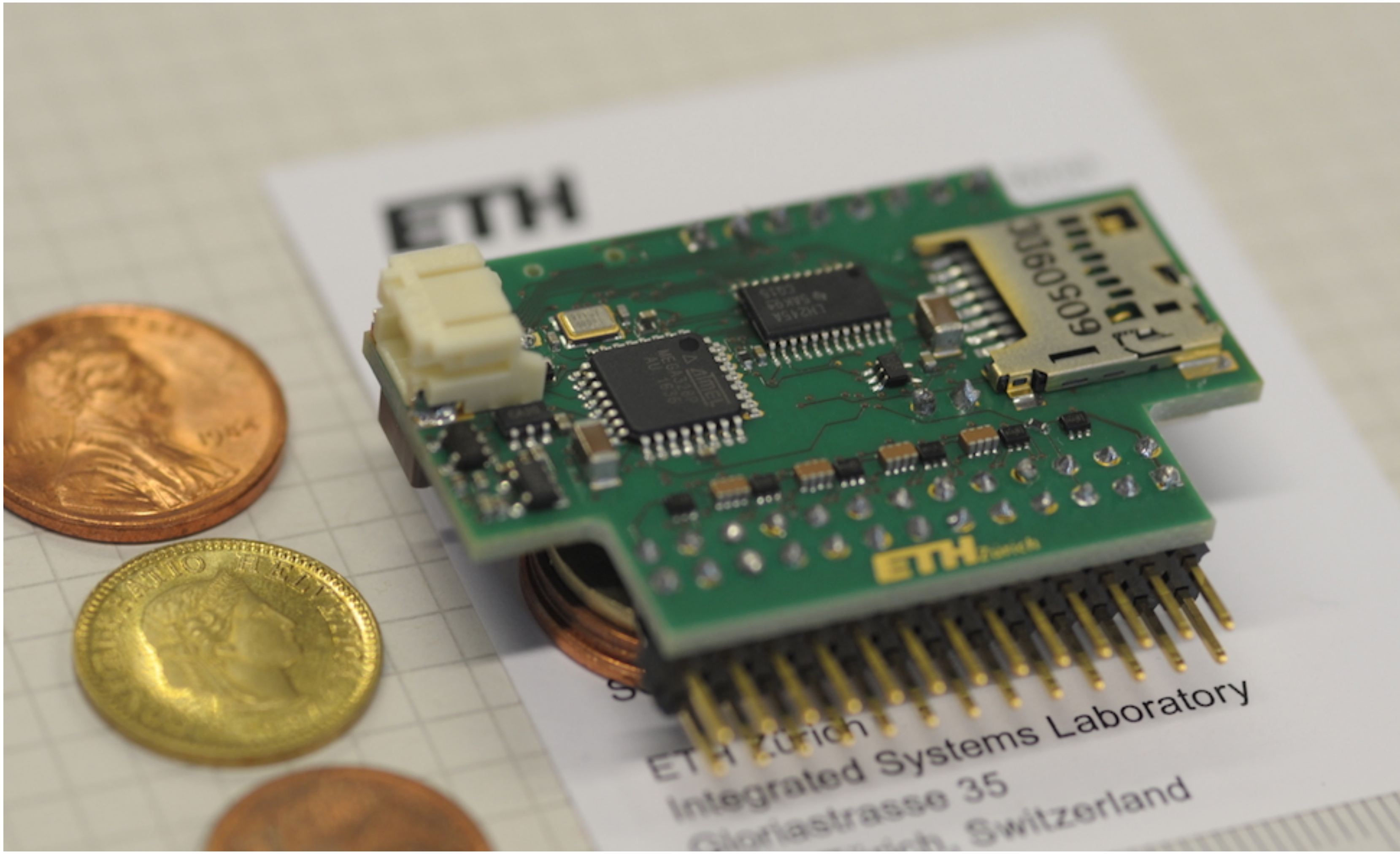


Combinational Logic Circuits and Design

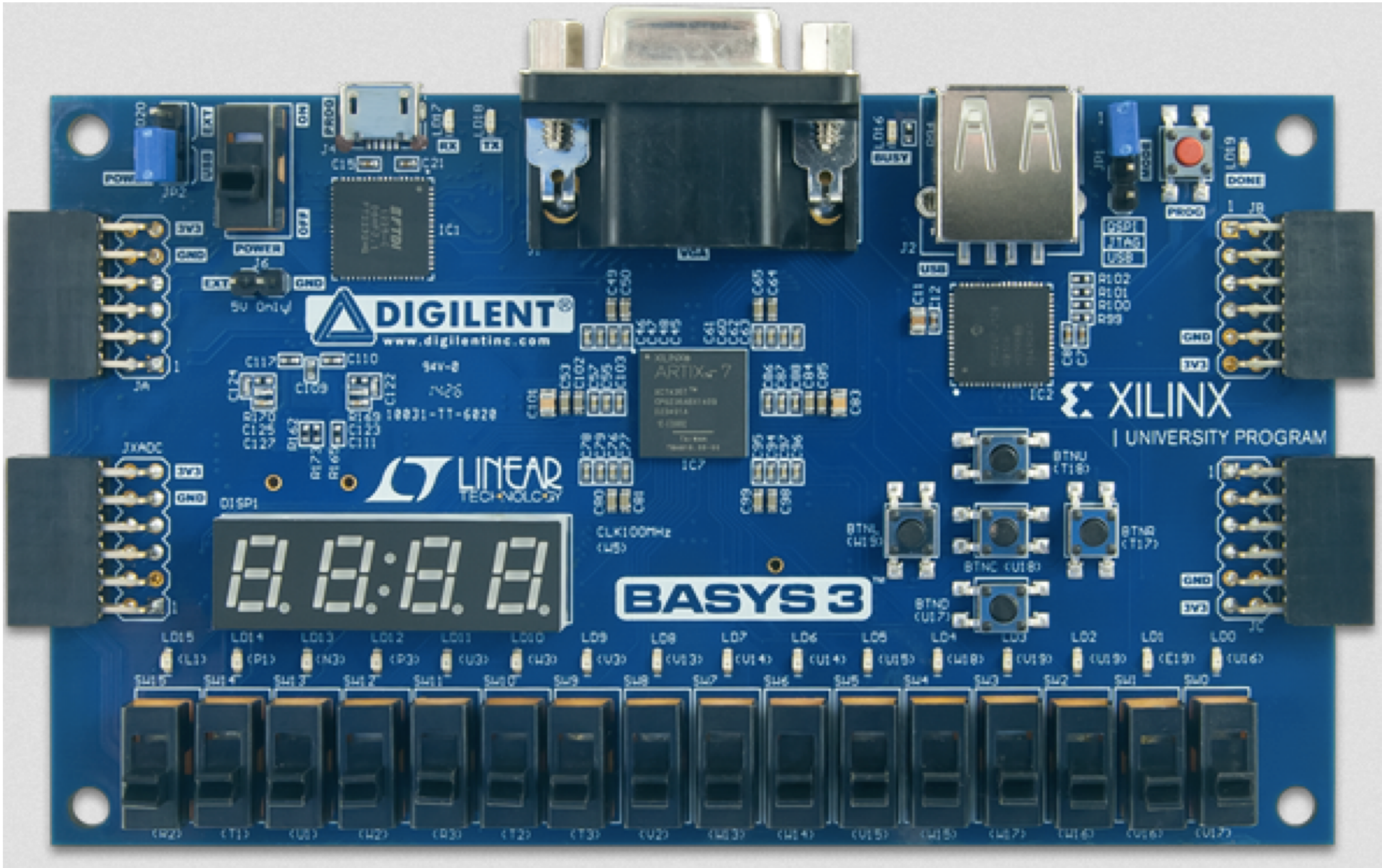
What Will We Learn Today?

- Building blocks of modern computers
  - Transistors
  - Logic gates
- Boolean algebra
- Combinational circuits
- How to use Boolean algebra to represent combinational circuits
- Minimizing logic circuits (if time permits)

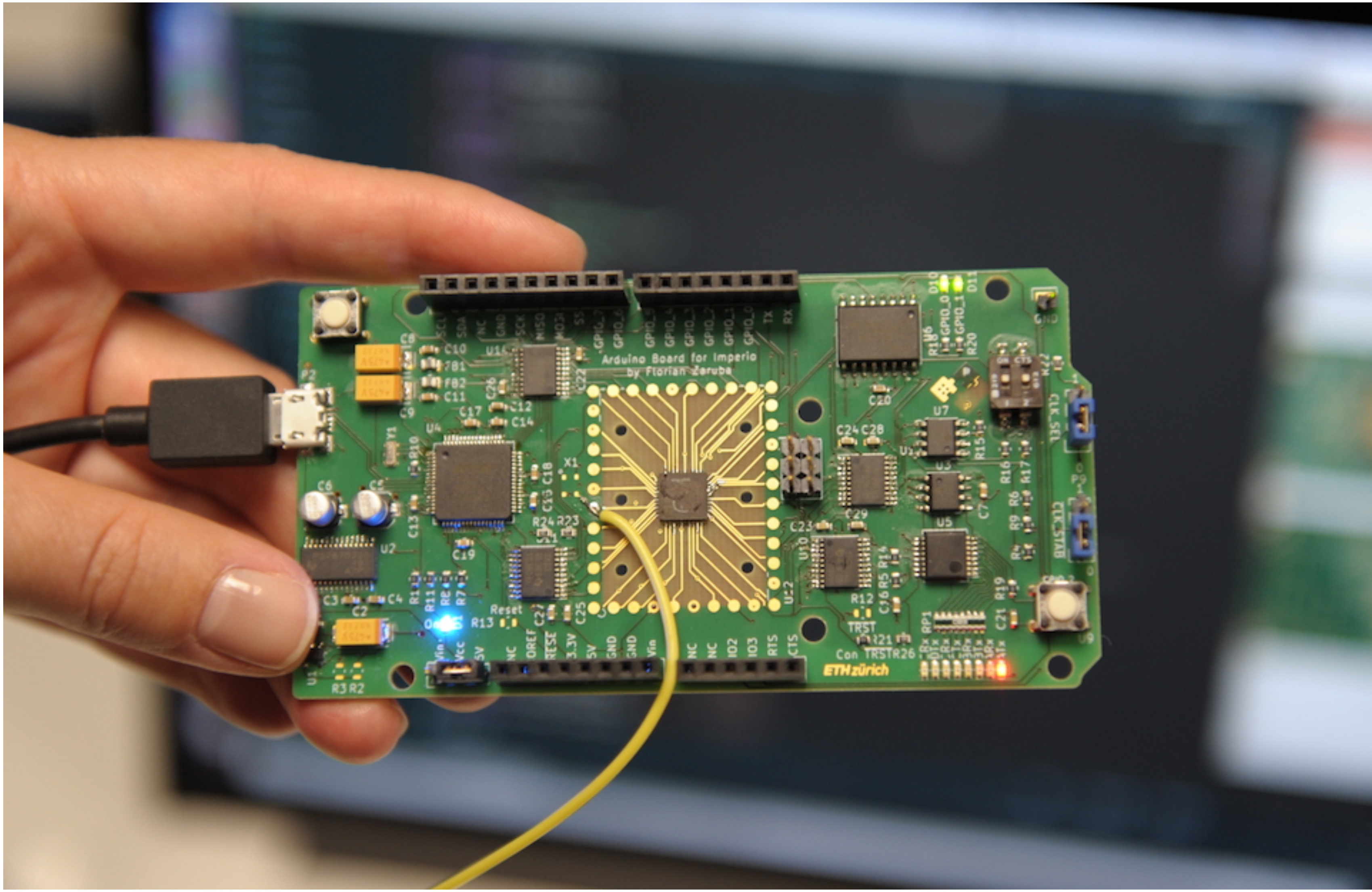
(Micro)-Processors



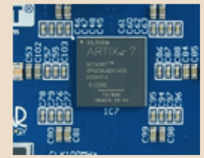
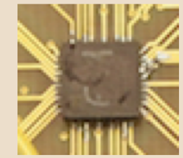
FPGAs


DIGILENT
www.digilentinc.com
XILINX
| UNIVERSITY PROGRAM
LINEAR
TECHNOLOGY
BASYS 3
CLK100MHz
POWER
BTNU
BTNL
BTNR
BTNC
BTND
DONE
PROG

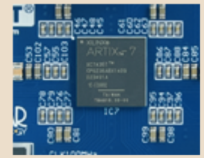
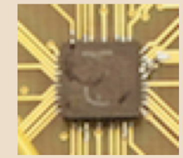
Custom ASICs



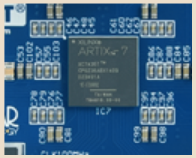
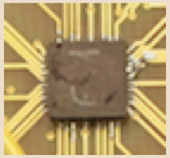
They All Look the Same

	Microprocessors	FPGAs	ASICs
			
In short:	Common building block of computers	Reconfigurable hardware, flexible	You customize everything

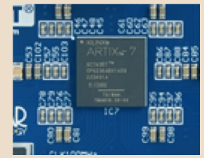
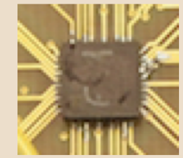
They All Look the Same

	Microprocessors	FPGAs	ASICs
			
In short:	Common building block of computers	Reconfigurable hardware, flexible	You customize everything
Program Development Time	minutes	days	months

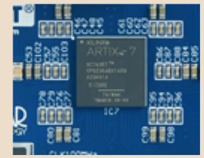
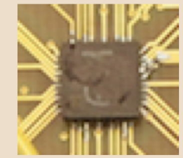
They All Look the Same

	Microprocessors	FPGAs	ASICs
			
In short:	Common building block of computers	Reconfigurable hardware, flexible	You customize everything
Program Development Time	minutes	days	months
Performance	o	+	++

They All Look the Same

	Microprocessors	FPGAs	ASICs
			
In short:	Common building block of computers	Reconfigurable hardware, flexible	You customize everything
Program Development Time	minutes	days	months
Performance	o	+	++
Good for	Ubiquitous Simple to use	Prototyping Small volume	Mass production, Max performance

They All Look the Same

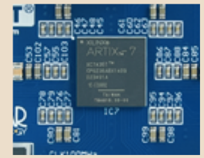
	Microprocessors	FPGAs	ASICs
			
In short:	Common building block of computers	Reconfigurable hardware, flexible	You customize everything
Program Development Time	minutes	days	months
Performance	o	+	++
Good for	Ubiquitous Simple to use	Prototyping Small volume	Mass production, Max performance
Programming	Executable file	Bit file	Design masks
Languages	C/C++/Java/…	Verilog/VHDL	Verilog/VHDL
Main Companies	Intel, ARM, AMD, Apple, NVIDIA	Xilinx, Altera	TSMC, Globalfoundries

They All Look the Same

Want to learn how these work

By programming these

Using this language

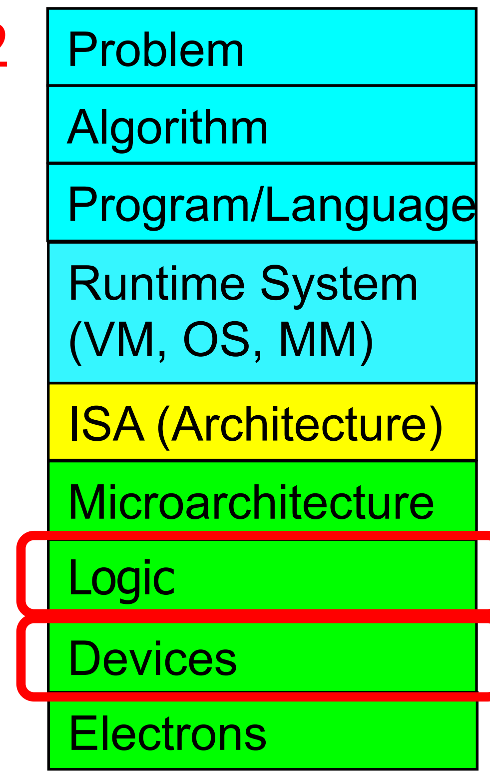
	Microprocessors	FPGAs	
			
In short	Common building block of computers	Reconfigurable hardware, flexible	nize ng
Program Development Time	minutes	days	months
Performance	o	+	++
Good for	Ubiquitous Simple to use	Prototyping	Mass production,
Programming	Executable file		
Languages	C/C++/Java/…	**Verilog/VHDL**	**Verilog/VHDL**
Main Companies	Intel, ARM, AMD, Apple, NVIDIA	Xilinx, Altera	TSMC, Globalfoundries

Building Blocks of Modern Computers

Transistors

Transistors

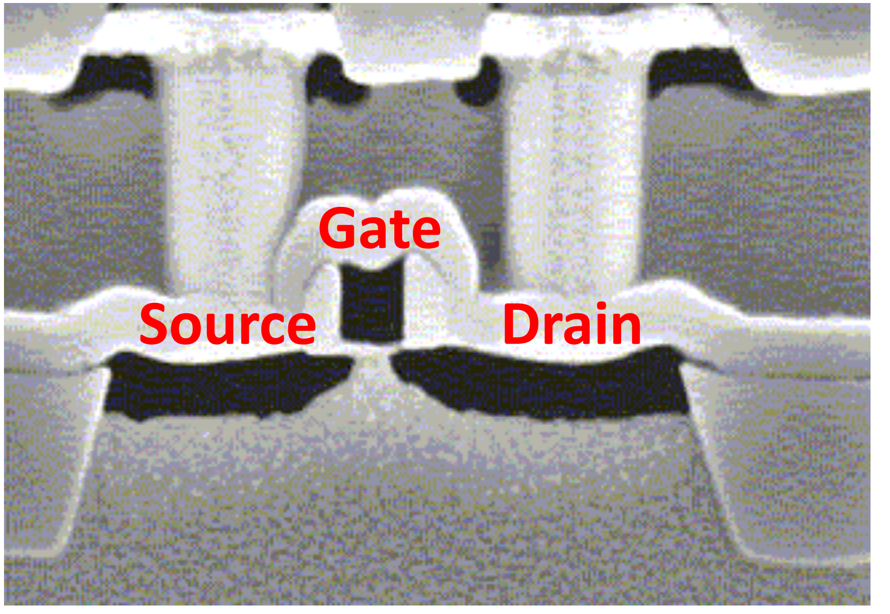
- **Computers are built from very large numbers of very simple structures**
  - Intel’s Pentium IV microprocessor, first offered for sale in 2000, was made up of more than 42 million MOS transistors
  - Intel’s Core i7 Broadwell-E, offered for sale in 2016, is made up of more than 3.2 billion MOS transistors



- **This lecture**
  - How the MOS transistor works (as a logic element)
  - How these transistors are connected to form logic gates
  - How logic gates are interconnected to form larger units that are needed to construct a computer

MOS Transistor

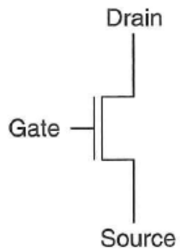
- By combining
  - Conductors (**M**etal)
  - Insulators (**O**xide)
  - **S**emiconductors



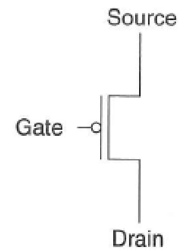
- We get a Transistor (MOS)
- Why is this useful?
  - We can combine many of these to realize simple logic gates
- The electrical properties of metal-oxide semiconductors are well beyond the scope of what we want to understand in this course
  - They are below our lowest level of abstraction

Different Types of MOS Transistors

- There are two types of MOS transistors: n-type and p-type



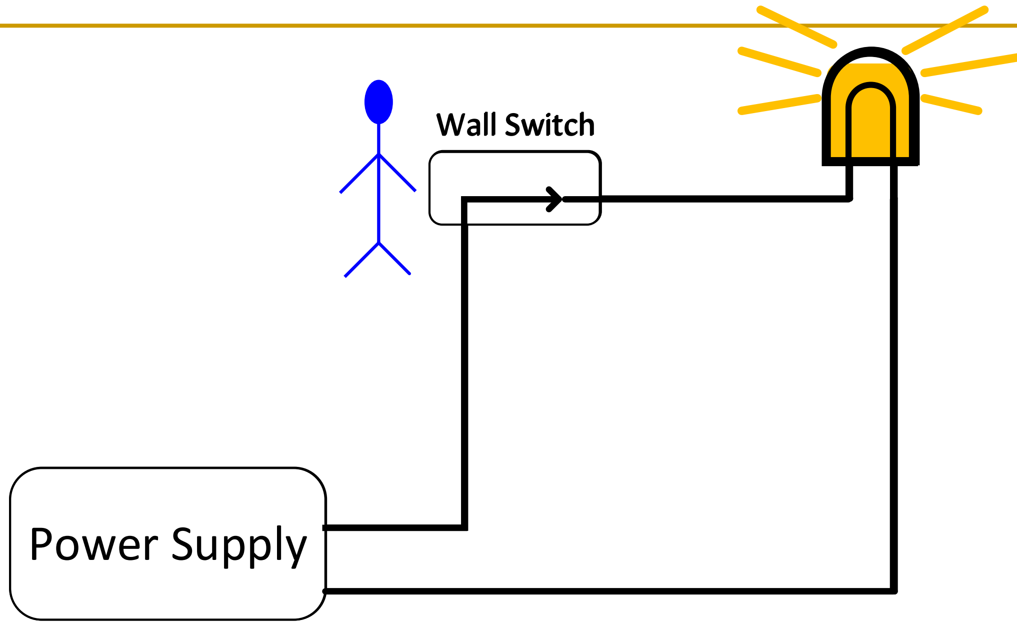
n-type



p-type

- They both operate "logically," very similar to the way wall switches work

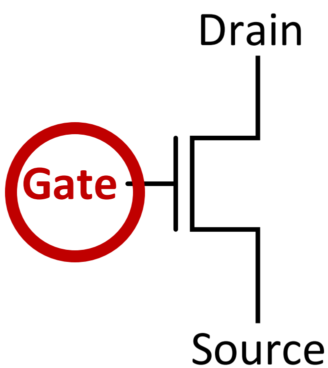
How Does a Transistor Work?



- In order for the lamp to glow, electrons must flow
- In order for electrons to flow, there must be a closed circuit from the power supply to the lamp and back to the power supply
- The lamp can be turned on and off by simply manipulating the wall switch to make or break the closed circuit

How Does a Transistor Work?

- Instead of the wall switch, we could use an n-type or a p-type MOS transistor to make or break the closed circuit



Schematic of an n-type MOS transistor

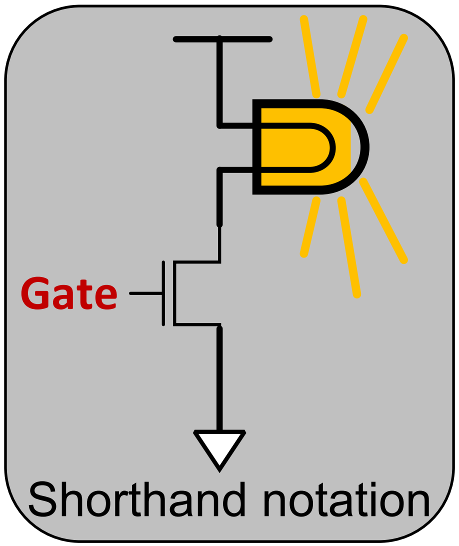
If the gate of an n-type transistor is supplied with a **high** voltage, the connection from source to drain acts like a piece of wire

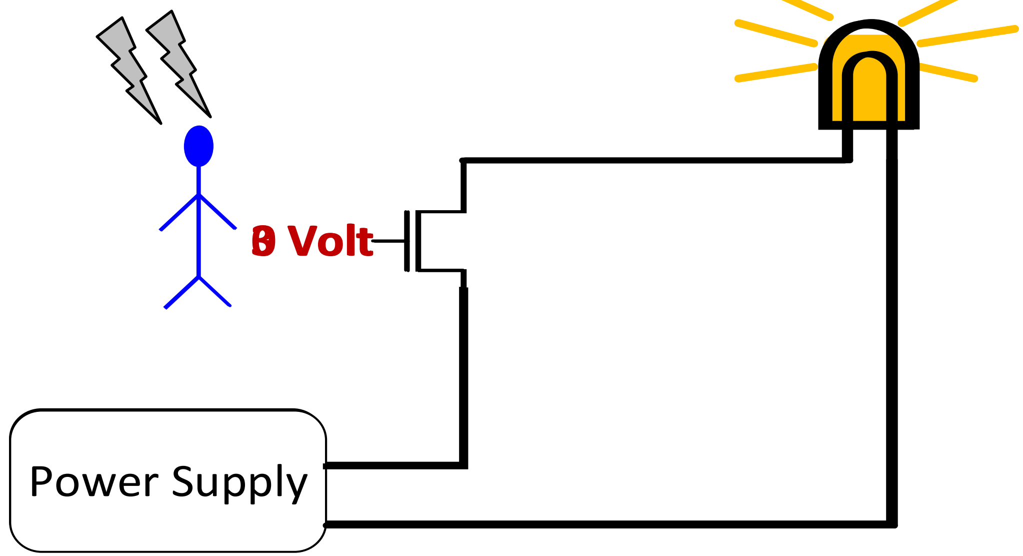
Depending on the technology, 0.3V to 3V

If the gate of the n-type transistor is supplied with 0V, the connection between the source and drain is broken

How Does a Transistor Work?

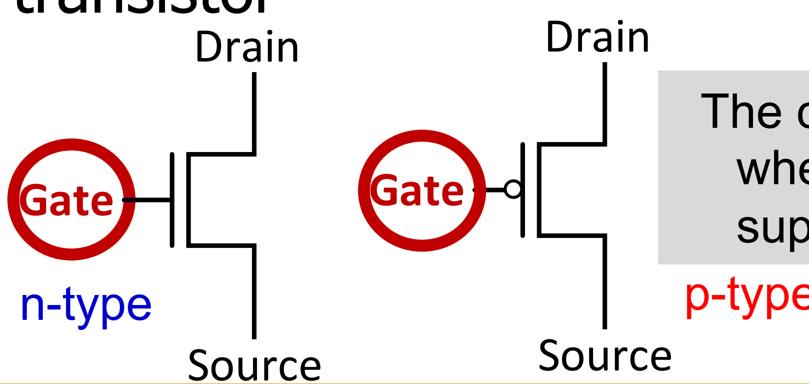
- The n-type transistor in a circuit with a battery and a bulb





- The p-type transistor works in exactly the opposite fashion from the n-type transistor

The circuit is closed when the gate is supplied with 3V

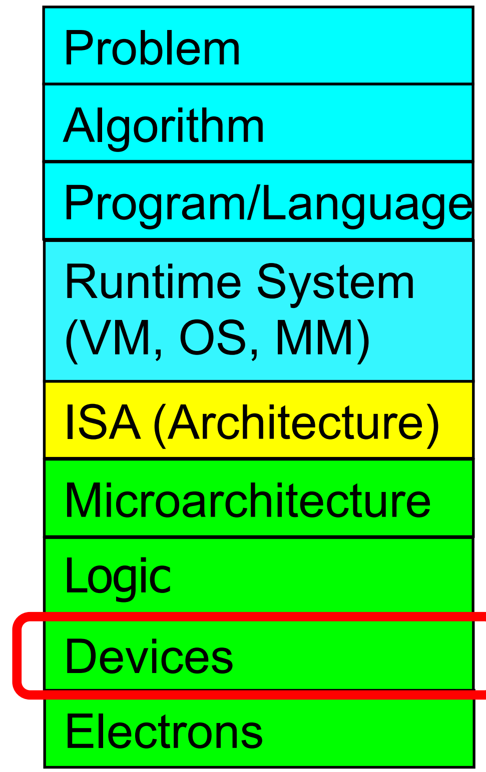


The circuit is closed when the gate is supplied with 0V

Logic Gates

One Level Higher in the Abstraction

- **Now, we know how a MOS transistor works**
- How do we build logic out of MOS transistors?
- We construct basic logic structures out of individual MOS transistors
- These logical units are named logic gates
  - They implement simple Boolean functions

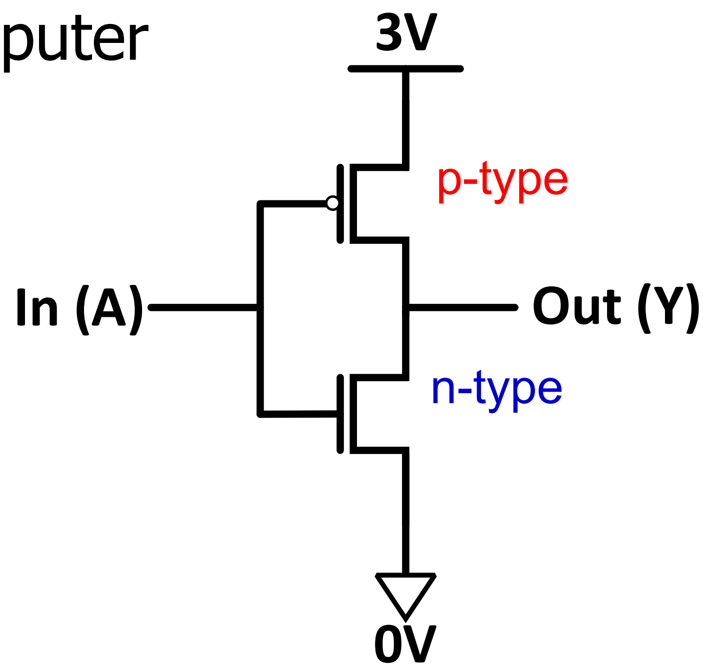


Making Logic Blocks Using CMOS Technology

- Modern computers use both n-type and p-type transistors, i.e. Complementary MOS (CMOS) technology

nMOS + pMOS = CMOS

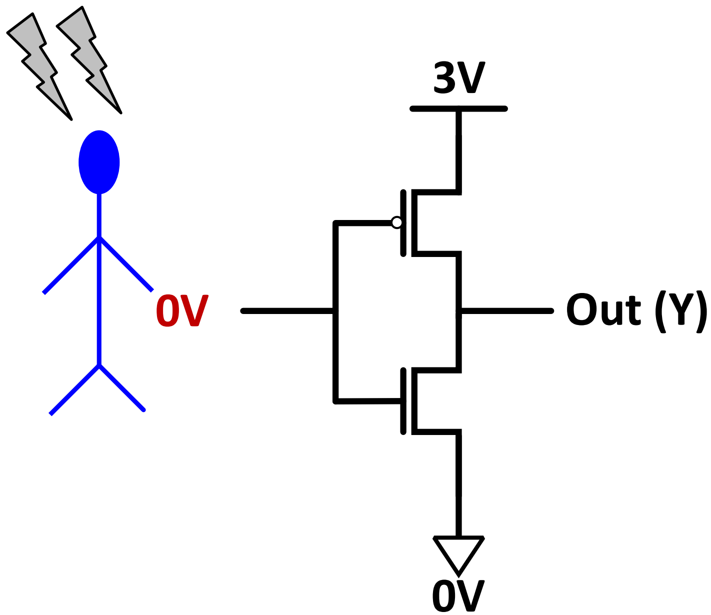
- The simplest logic structure that exists in a modern computer

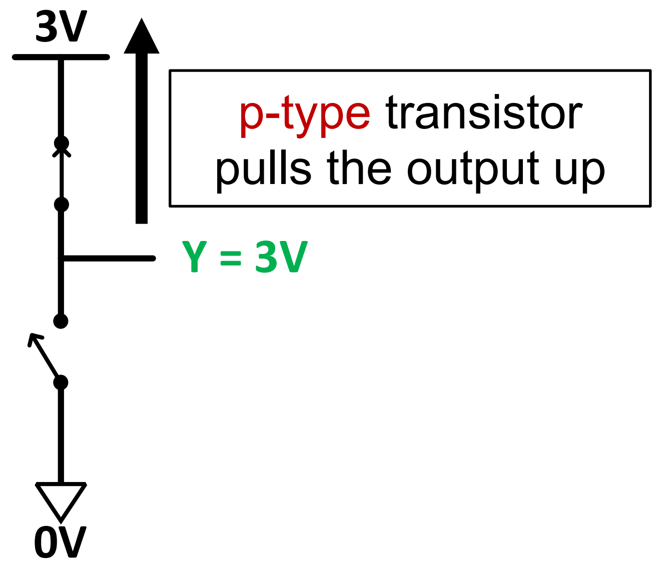


What does this circuit do?

Functionality of Our CMOS Circuit

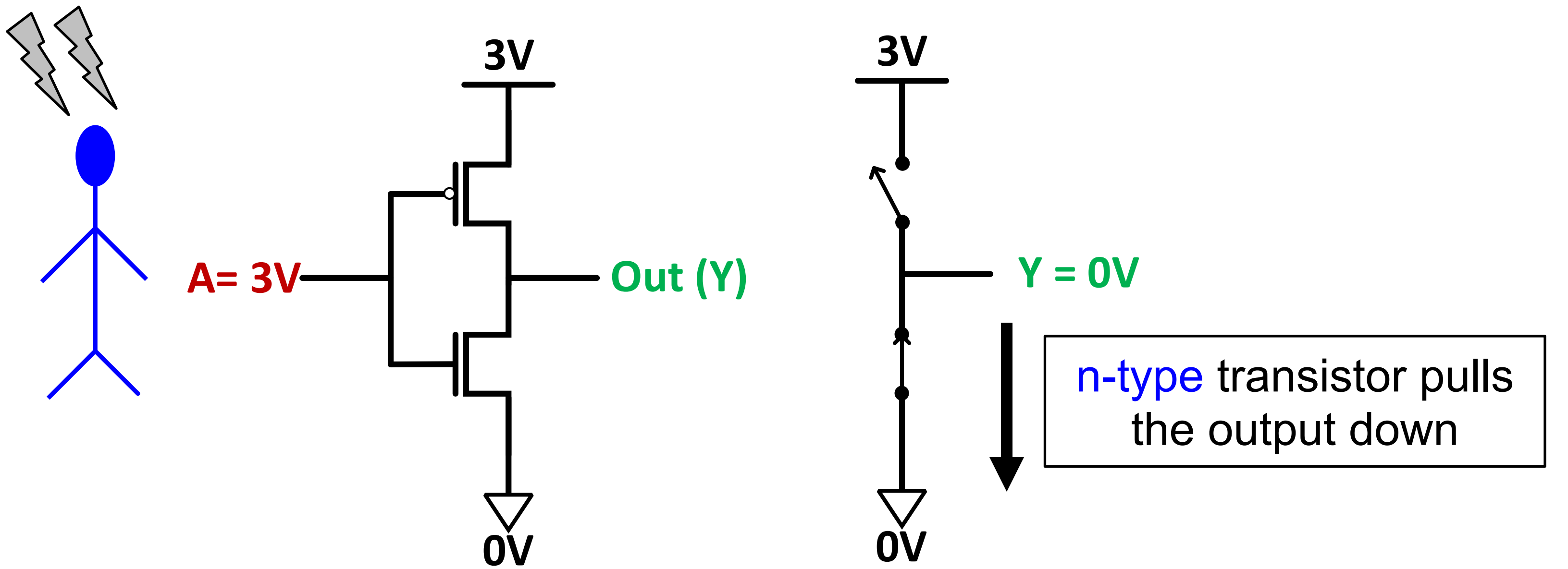
What happens when the input is connected to 0V?





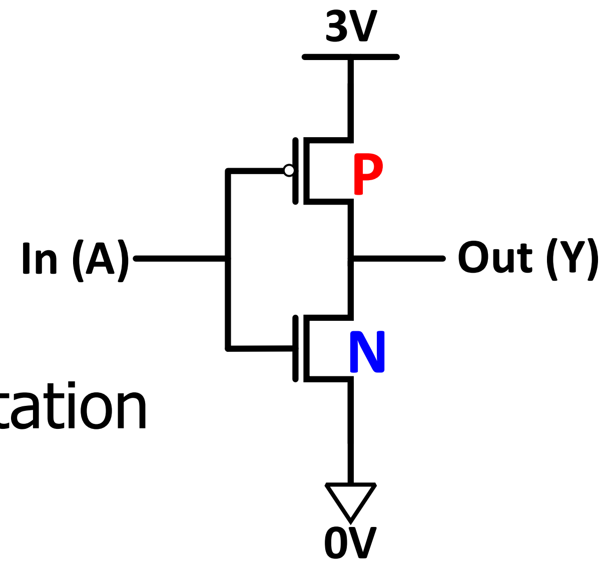
Functionality of Our CMOS Circuit

What happens when the input is connected to 3V?



CMOS NOT Gate

- This is actually the CMOS NOT Gate
- Why do we call it NOT?
  - If A = 0V then Y = 3V
  - If A = 3V then Y = 0V
- **Digital circuit:** one possible interpretation
  - Interpret 0V as logical (binary) 0 value
  - Interpret 3V as logical (binary) 1 value

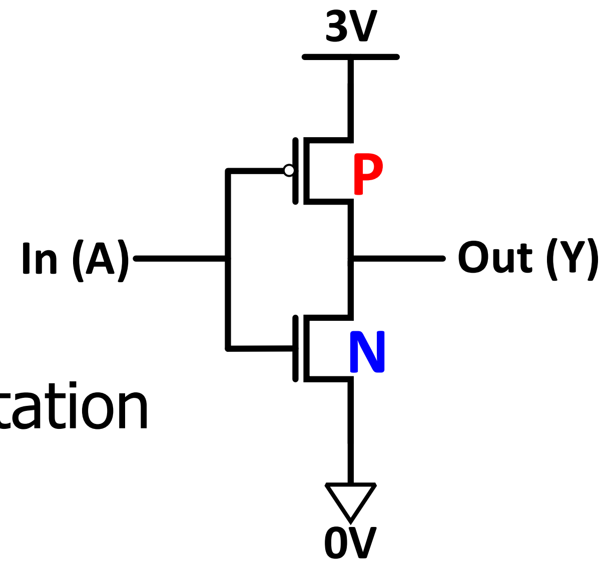


A	P	N	Y
0	ON	OFF	1
1	OFF	ON	0

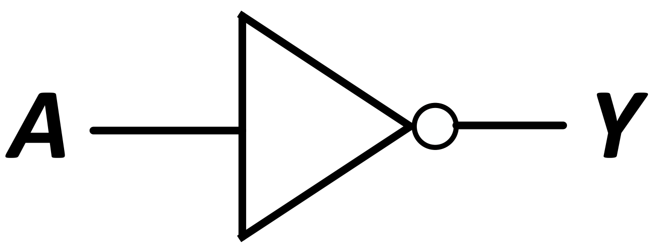
$$Y = \bar{A}$$

CMOS NOT Gate

- This is actually the CMOS NOT Gate
- Why do we call it NOT?
  - If A = 0V then Y = 3V
  - If A = 3V then Y = 0V
- **Digital circuit:** one possible interpretation
  - Interpret 0V as logical (binary) 0 value
  - Interpret 3V as logical (binary) 1 value



$$Y = \bar{A}$$



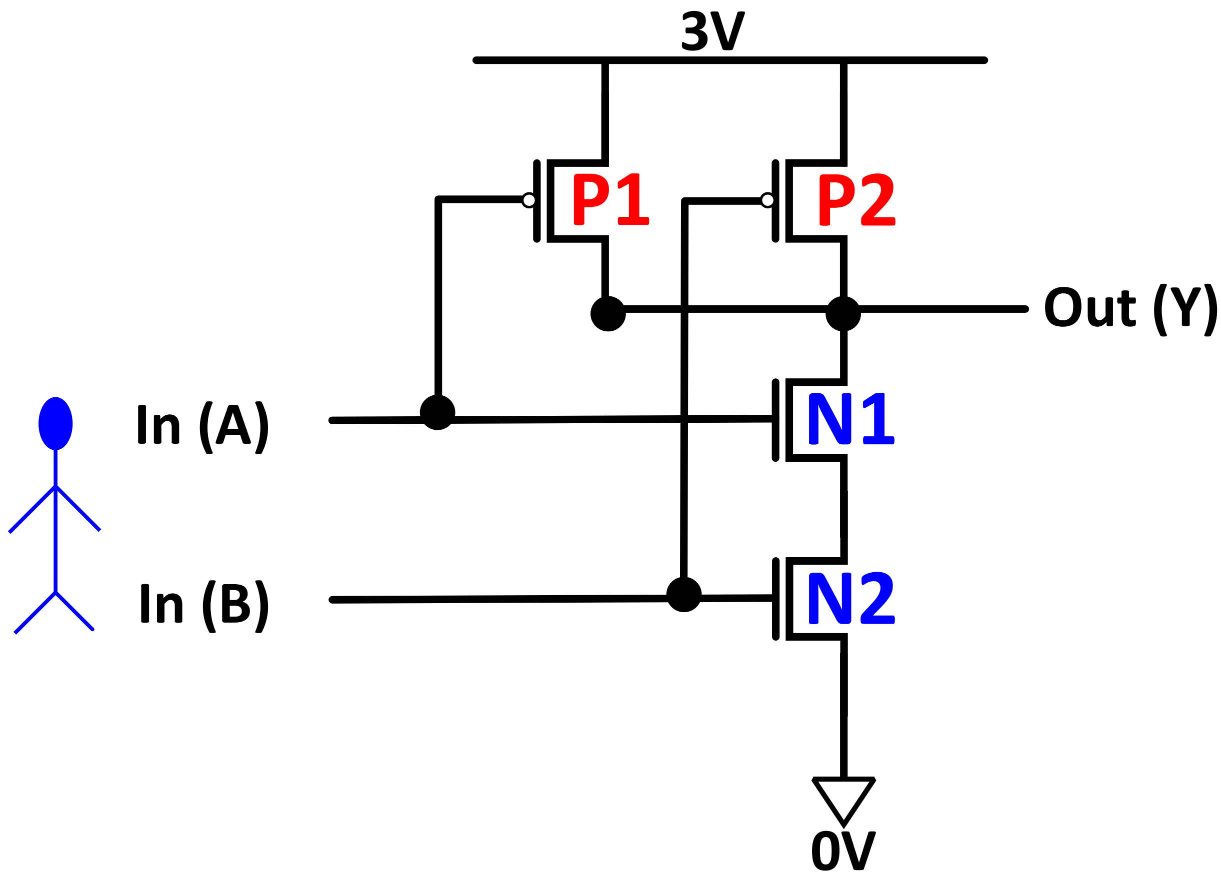
We call it a NOT gate
or an inverter

Truth table: shows what is the logical output of the circuit for each possible input

A	*Y*
0	1
1	0

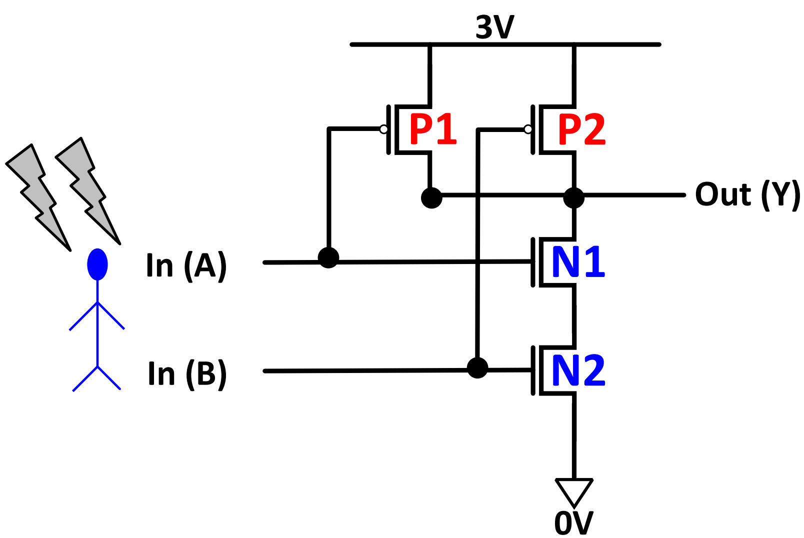
Another CMOS Gate: What Is This?

- Let's build more complex gates!



CMOS NAND Gate

- Let's build more complex gates!



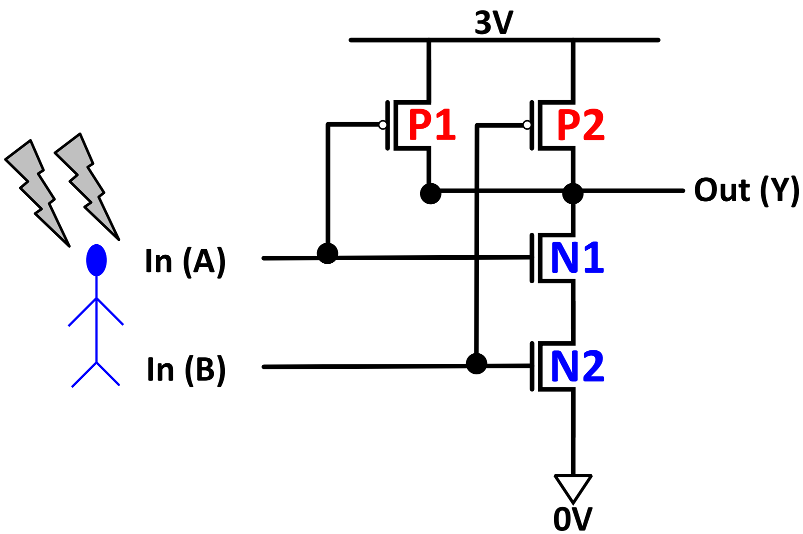
$$Y = \overline{A \cdot B} = \overline{AB}$$

A	B	P1	P2	N1	N2	Y
0	0	ON	ON	OFF	OFF	1
0	1	ON	OFF	OFF	ON	1
1	0	OFF	ON	ON	OFF	1
1	1	OFF	OFF	ON	ON	0

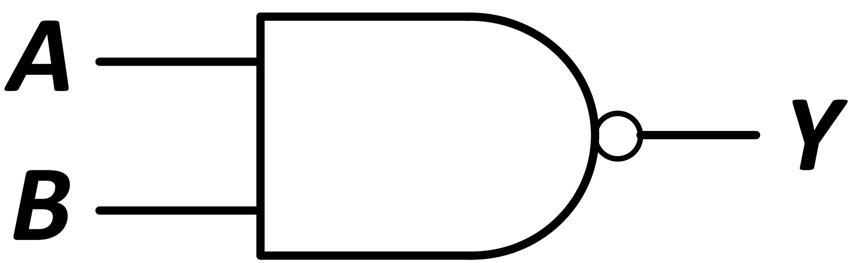
  - P1 and P2 are in parallel; only one must be ON to pull the output up to 3V
  - N1 and N2 are connected in series; both must be ON to pull the output to 0V

CMOS NAND Gate

- Let's build more complex gates!



$$Y = \overline{A \cdot B} = \overline{AB}$$



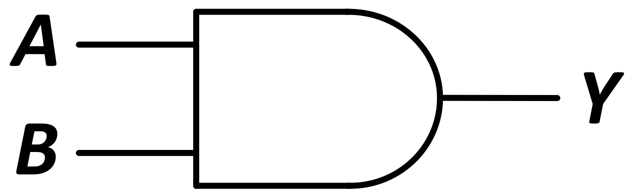
A	B	Y
0	0	1
0	1	1
1	0	1
1	1	0

CMOS AND Gate

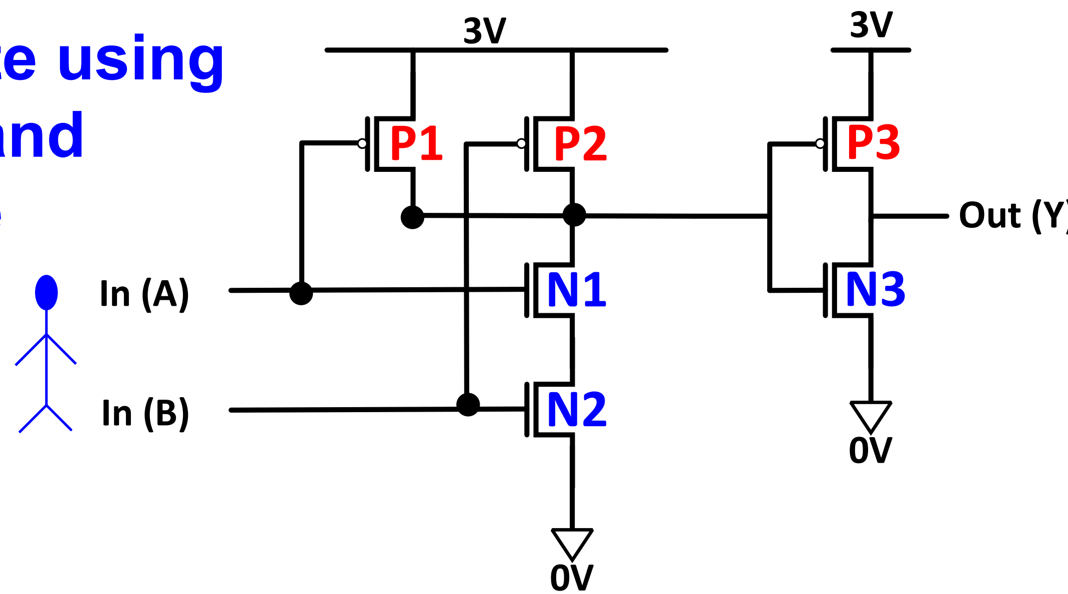
- How can we make an AND gate?

A	B	Y
0	0	0
0	1	0
1	0	0
1	1	1

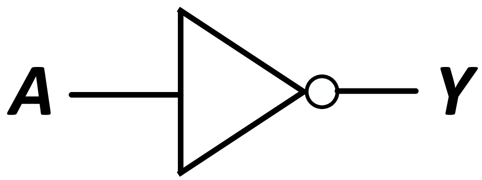
$$Y = A \cdot B = AB$$

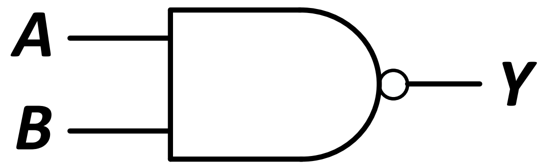


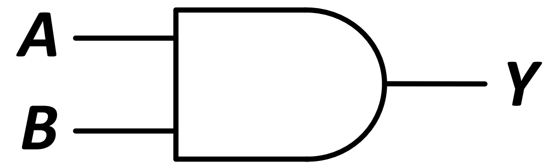
We make an AND gate using one NAND gate and one NOT gate



CMOS NOT, NAND, AND Gates



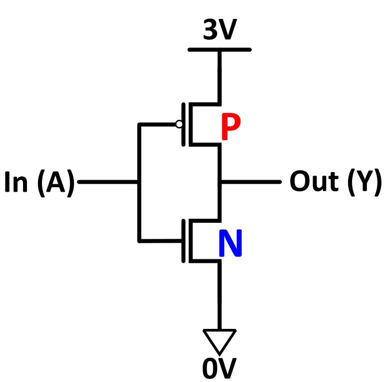


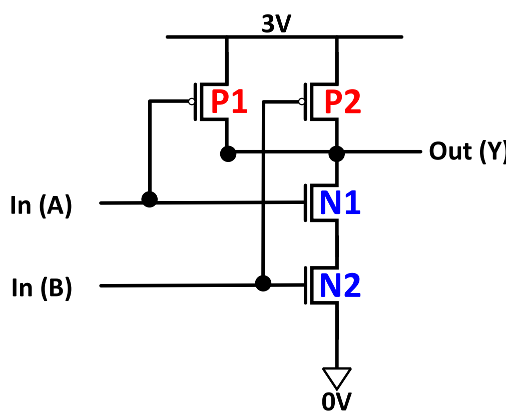


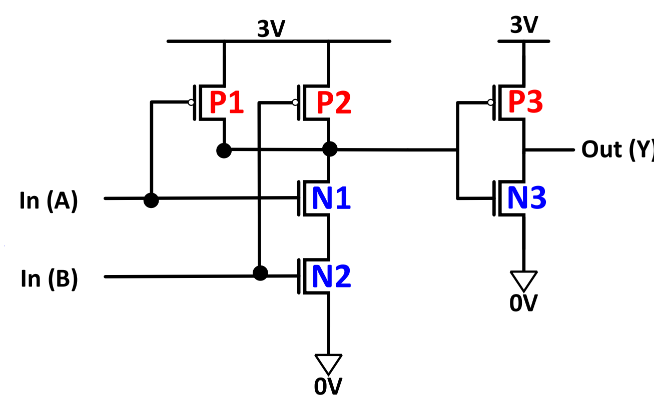
A	Y
0	1
1	0

A	B	Y
0	0	1
0	1	1
1	0	1
1	1	0

A	B	Y
0	0	0
0	1	0
1	0	0
1	1	1

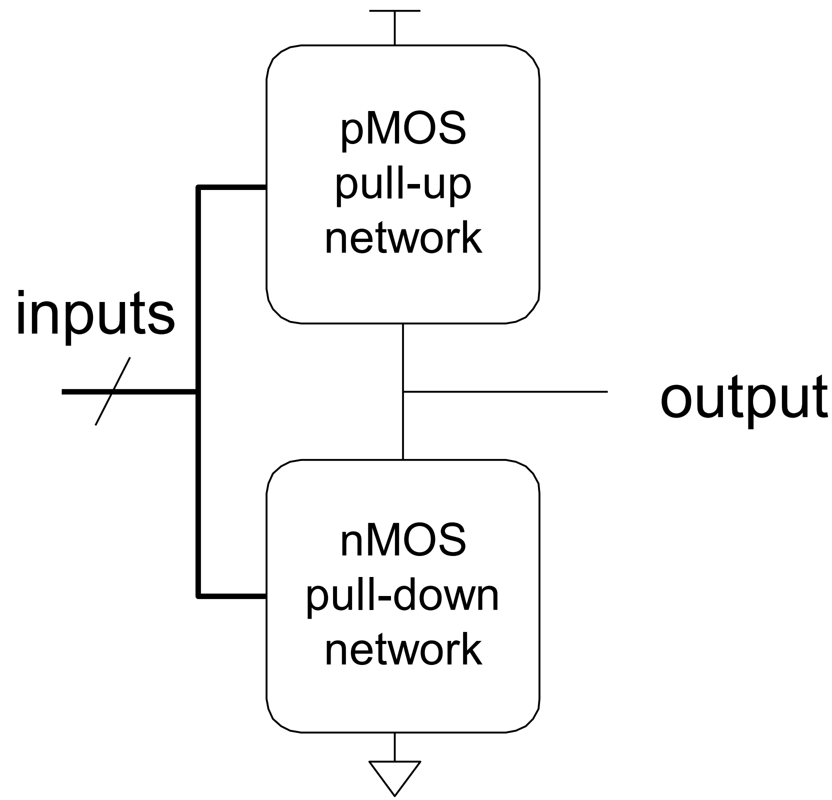






General CMOS Gate Structure

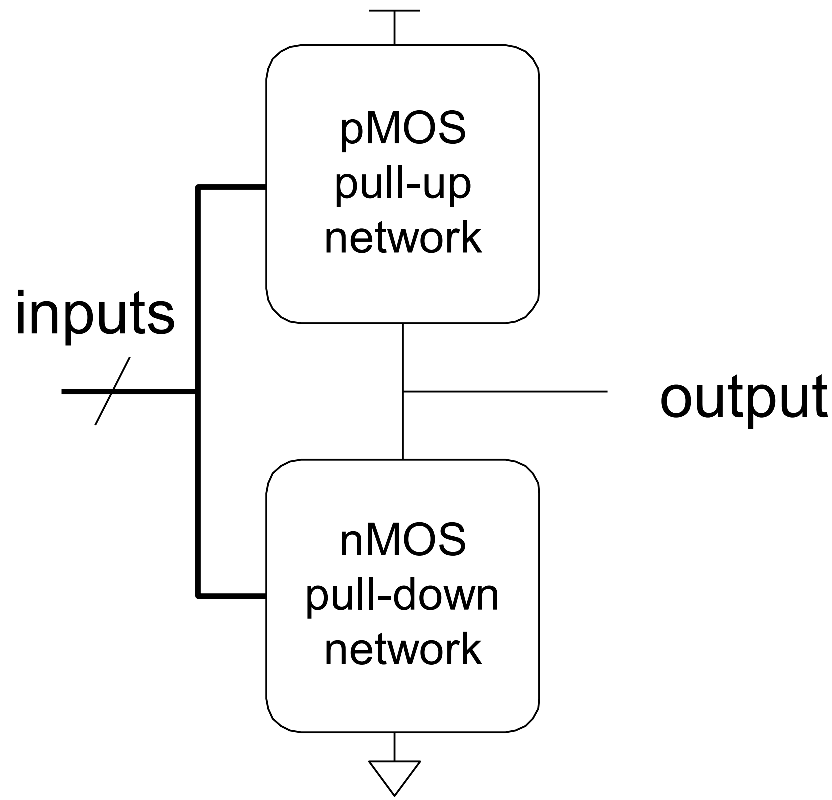
- The general form used to construct any inverting logic gate, such as: NOT, NAND, or NOR
  - The networks may consist of transistors in series or in parallel
  - When transistors are in parallel, the network is **ON** if one of the transistors is **ON**
  - When transistors are in series, the network is **ON** only if all transistors are **ON**



pMOS transistors are used for pull-up
nMOS transistors are used for pull-down

General CMOS Gate Structure (II)

- Exactly one network should be ON, and the other network should be OFF at any given time
  - If both networks are ON at the same time, there is a short circuit → likely incorrect operation
  - If both networks are OFF at the same time, the output is floating → undefined



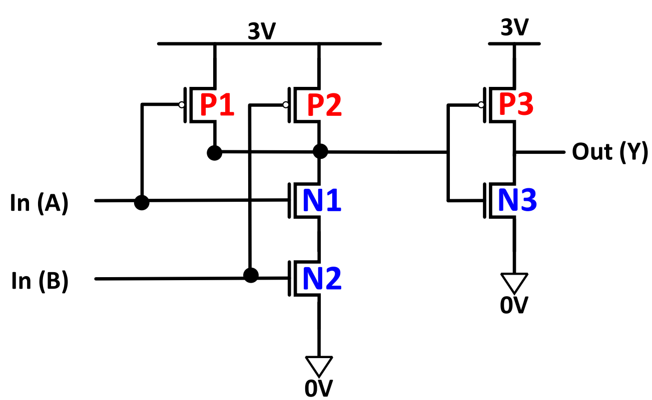
pMOS transistors are used for pull-up
nMOS transistors are used for pull-down

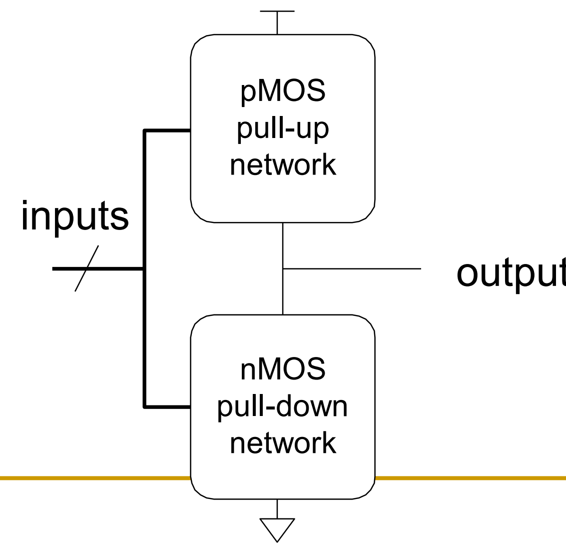
Digging Deeper: Why This Structure?

- MOS transistors are **not perfect** switches

- pMOS transistors pass 1's well but 0's poorly
- nMOS transistors pass 0's well but 1's poorly

- pMOS transistors are good at "pulling up" the output
- nMOS transistors are good at "pulling down" the output





See Section 1.7 in H&H

Digging Deeper: Latency

- Which one is faster?
  - Transistors in series
  - Transistors in parallel

- Series connections are slower than parallel connections
  - More resistance on the wire

- How do you alleviate this latency?
  - See H&H Section 1.7.8 for an example: pseudo-nMOS Logic

Digging Deeper: Power Consumption

- Dynamic Power Consumption
  - $C * V^2 * f$
    - C = capacitance of the circuit (wires and gates)
    - V = supply voltage
    - f = charging frequency of the capacitor
- Static Power consumption
  - $V * I_{leakage}$
    - supply voltage * leakage current
- Energy Consumption
  - Power * Time
- See more in H&H Chapter 1.8

Common Logic Gates

Buffer

A	Z
0	0
1	1

AND

A	B	Z
0	0	0
0	1	0
1	0	0
1	1	1

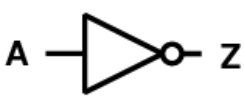
OR

A	B	Z
0	0	0
0	1	1
1	0	1
1	1	1

XOR

A	B	Z
0	0	0
0	1	1
1	0	1
1	1	0

Inverter



A	Z
0	1
1	0

NAND

A	B	Z
0	0	1
0	1	1
1	0	1
1	1	0

NOR

A	B	Z
0	0	1
0	1	0
1	0	0
1	1	0

XNOR

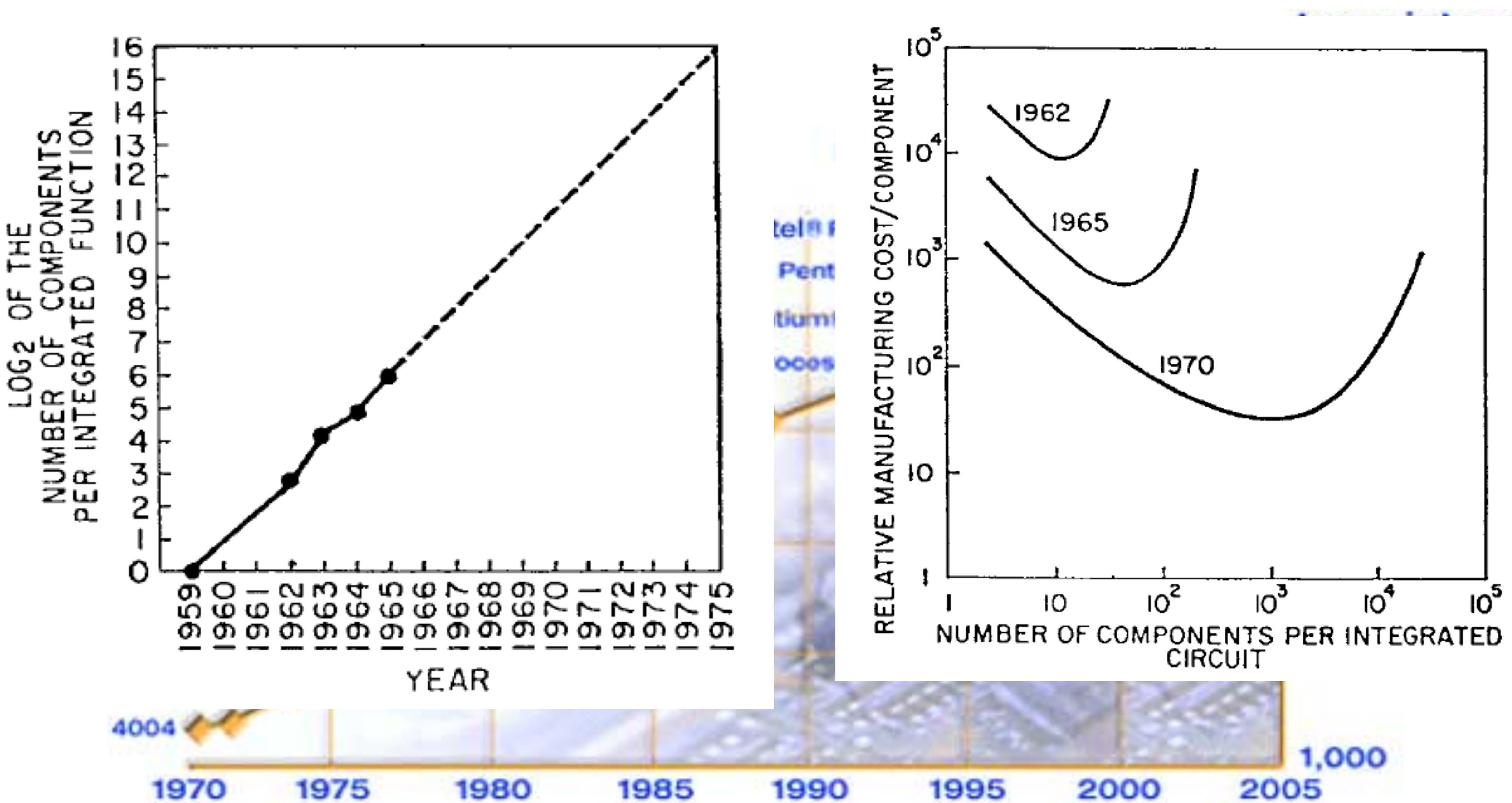
A	B	Z
0	0	1
0	1	0
1	0	0
1	1	1

Larger Gates

- We can extend the gates to more than 2 inputs
- Example: 3-input AND gate, 10-input NOR gate
- See your readings

Aside: Moore's Law:
Enabler of Many Gates on a Chip

An Enabler: Moore’s Law

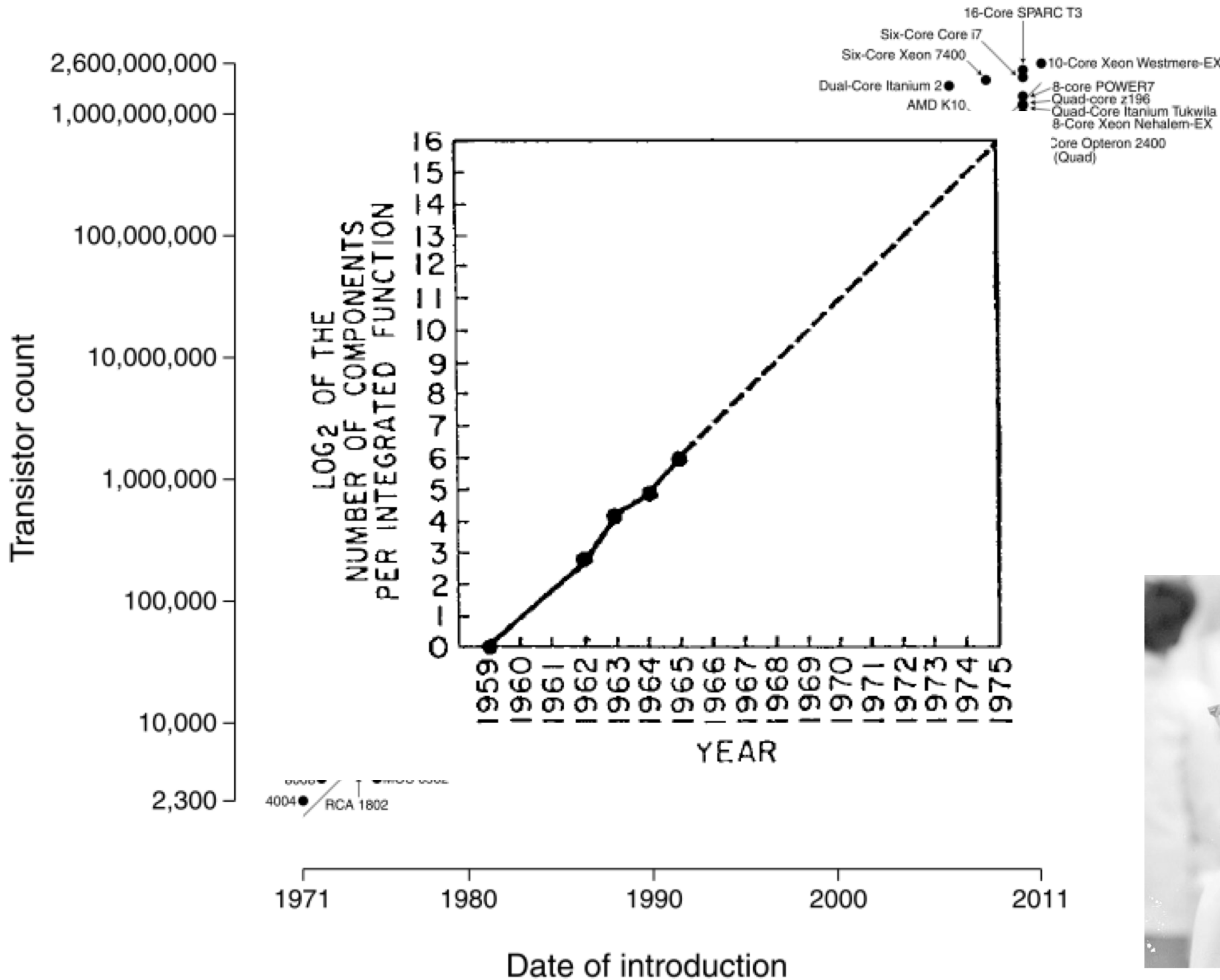


Moore, “Cramming more components onto integrated circuits,” Electronics Magazine, 1965.

Component counts double every other year

Image source: Intel

Microprocessor Transistor Counts 1971-2011 & Moore's Law



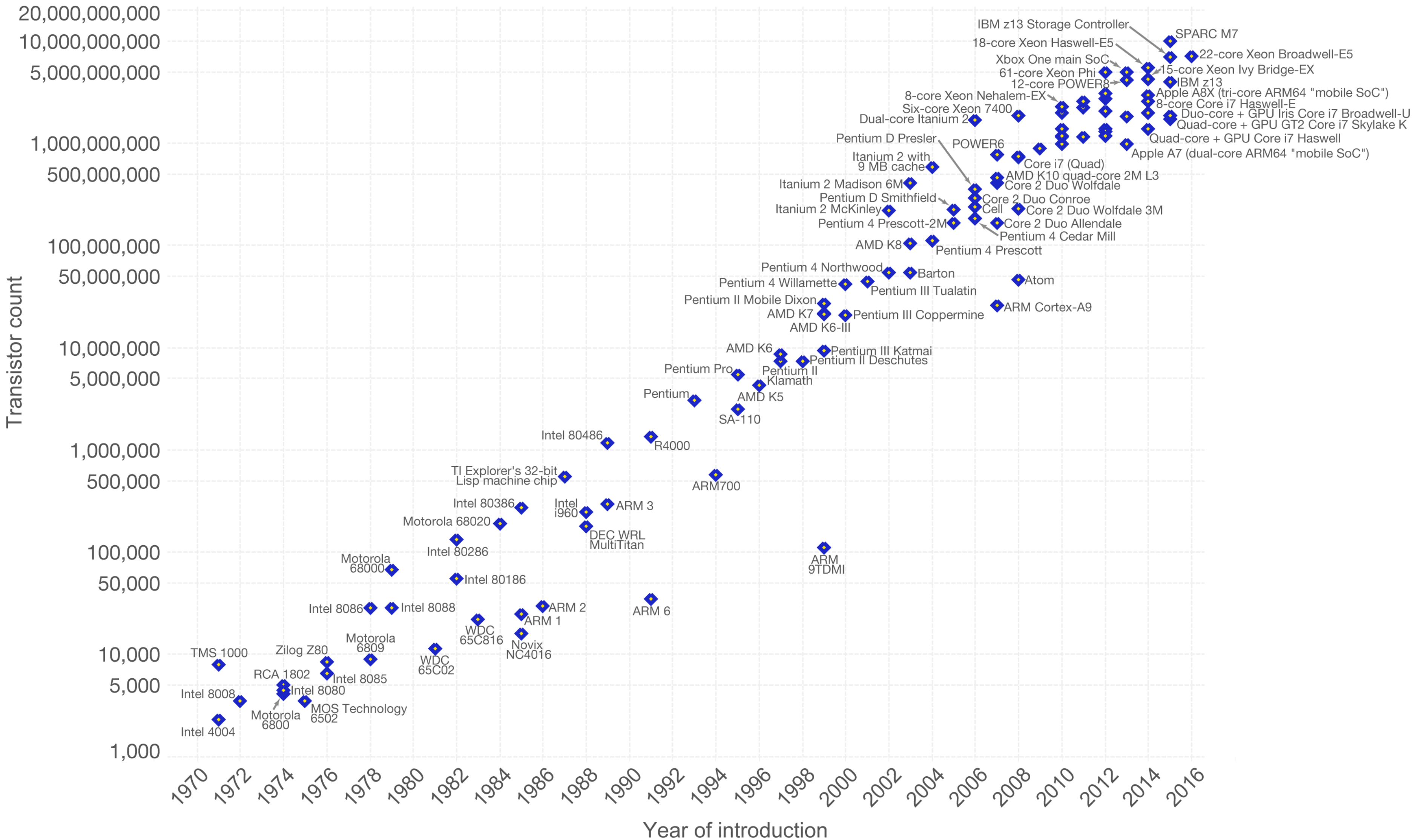
Number of transistors on an integrated circuit doubles ~ every two years

Image source: Wikipedia

Moore's Law – The number of transistors on integrated circuit chips (1971-2016)

Our World in Data

Moore's law describes the empirical regularity that the number of transistors on integrated circuits doubles approximately every two years. This advancement is important as other aspects of technological progress – such as processing speed or the price of electronic products – are strongly linked to Moore's law.

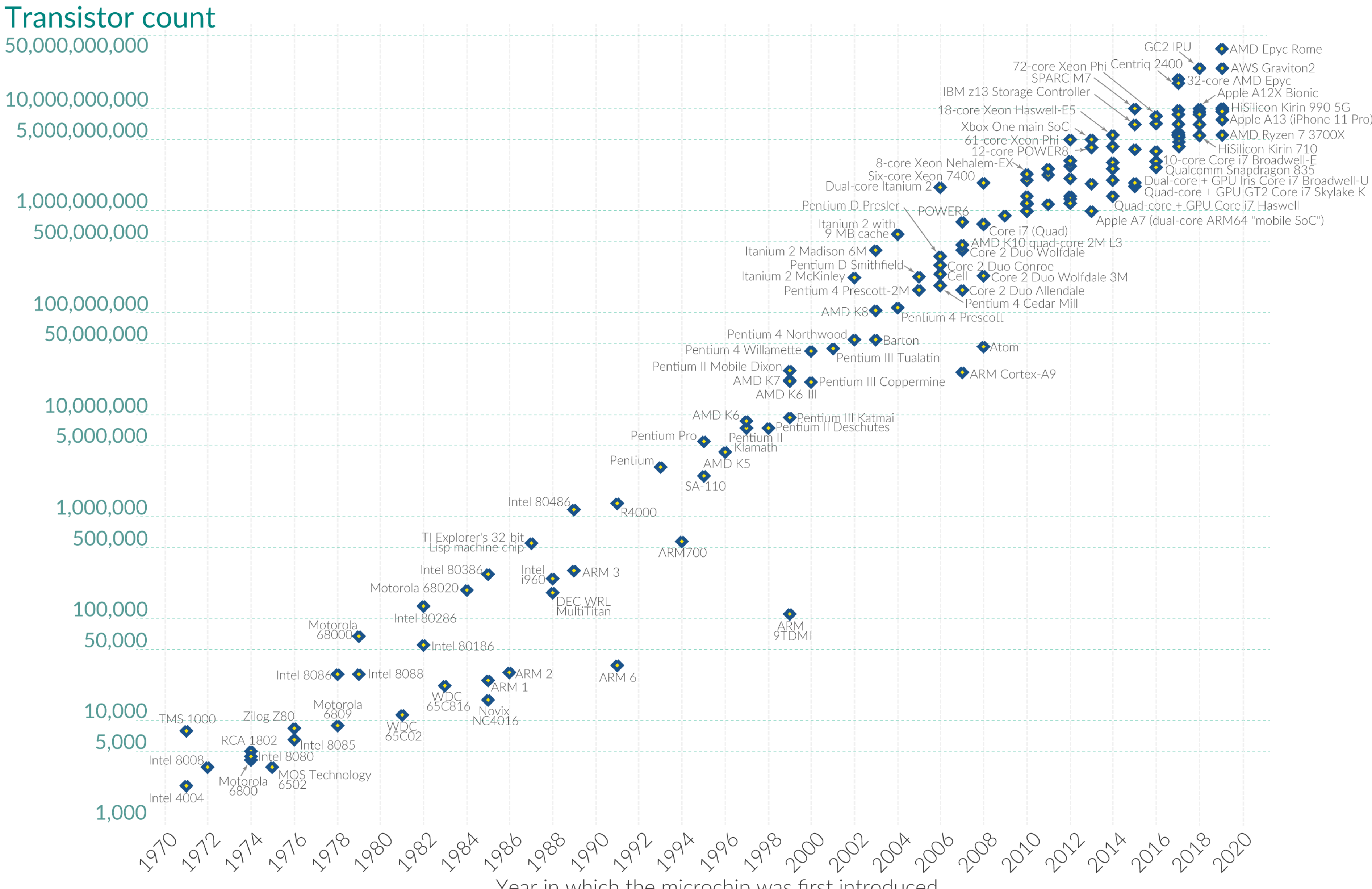


Data source: Wikipedia (https://en.wikipedia.org/wiki/Transistor_count)
The data visualization is available at OurWorldinData.org. There you find more visualizations and research on this topic.

Moore's Law: The number of transistors on microchips doubles every two years

Our World in Data

Moore's law describes the empirical regularity that the number of transistors on integrated circuits doubles approximately every two years. This advancement is important for other aspects of technological progress in computing – such as processing speed or the price of computers.



Data source: Wikipedia (wikipedia.org/wiki/Transistor_count)

OurWorldinData.org – Research and data to make progress against the world's largest problems.

Recommended Reading

- Moore, "Cramming more components onto integrated circuits," Electronics Magazine, 1965.

- Only 3 pages

- A quote:

  "*With unit cost falling as the number of components per circuit rises, by 1975 economics may dictate squeezing as many as 65 000 components on a single silicon chip.*"

- Another quote:

  "*Will it be possible to remove the heat generated by tens of thousands of components in a single silicon chip?*"

How Do We Keep Moore's Law

- **Manufacturing smaller transistors/structures**
  - Some structures are already a few atoms in size

- **Developing materials with better properties**
  - Copper instead of Aluminum (better conductor)
  - Hafnium Oxide, air for Insulators
  - Making sure all materials are compatible is the challenge

- **Optimizing the manufacturing steps**
  - How to use 193nm ultraviolet light to pattern 20nm structures

- **New technologies**
  - FinFET, Gate All Around transistor, Single Electron Transistor…

Combinational Logic Circuits

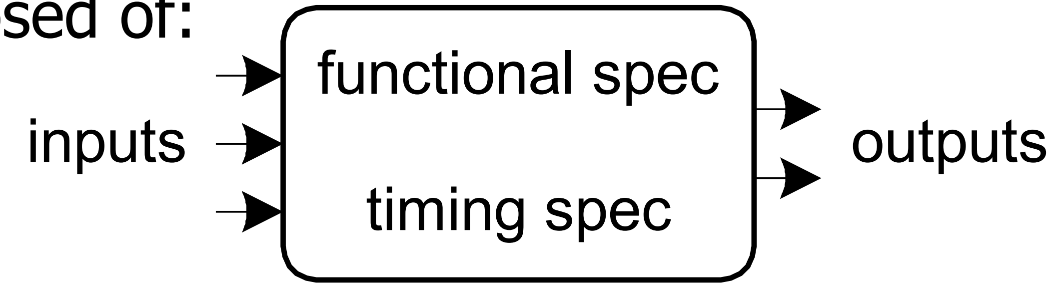
We Can Now Build Logic Circuits

Now, we understand the workings of the basic logic gates

What is our next step?

Build some of the logic structures that are important components of the microarchitecture of a computer!

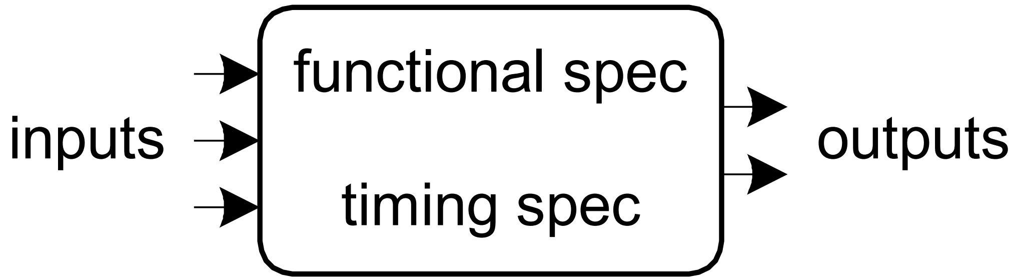
- A logic circuit is composed of:
  - Inputs
  - Outputs



- *Functional specification* (describes relationship between inputs and outputs)

- *Timing specification* (describes the delay between inputs changing and outputs responding)

Types of Logic Circuits



- **Combinational Logic**
  - Memoryless
  - Outputs are strictly dependent on the combination of input values that are being applied to circuit *right now*
  - In some books called Combinatorial Logic
- **Later we will learn: Sequential Logic**
  - Has memory
    - Structure stores history → Can "store" data values
  - Outputs are determined by previous (historical) and current values of inputs

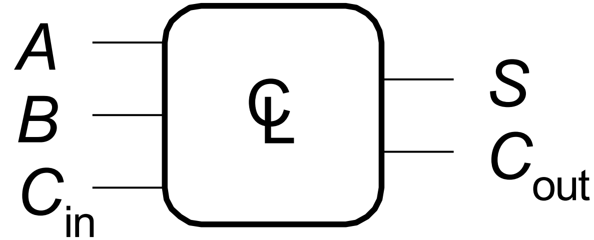
Boolean Equations

Functional Specification

- Functional specification of outputs in terms of inputs
- What do we mean by "function"?
  - Unique mapping from input values to output values
  - The same input values produce the same output value every time
  - No memory (does not depend on the history of input values)

- ***Example (full 1-bit adder – more later):***

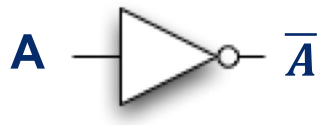
$$S = F(A, B, C_{in})$$
$$C_{out} = G(A, B, C_{in})$$



$$S = A \oplus B \oplus C_{in}$$
$$C_{out} = AB + AC_{in} + BC_{in}$$

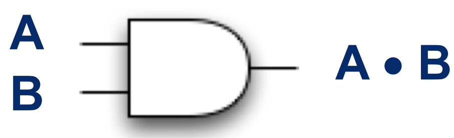
Simple Equations: NOT / AND / OR

$\overline{A}$ *(reads "not A")* is 1 iff A is 0



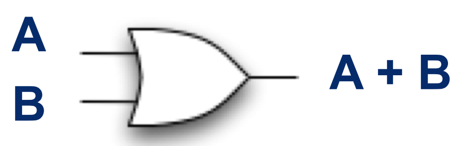
A	$\overline{A}$
0	1
1	0

A • B *(reads "A and B")* is 1 iff A and B are both 1



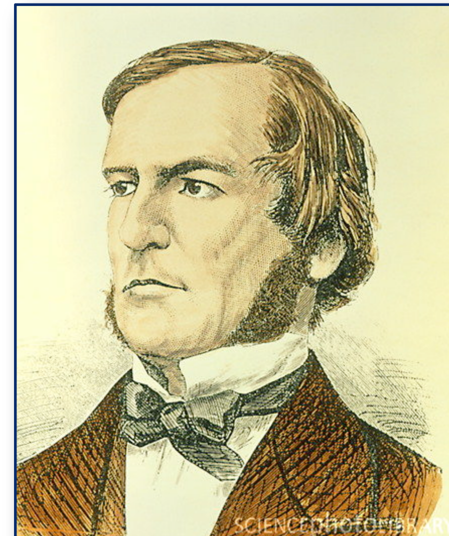
A	B	$A \bullet B$
0	0	0
0	1	0
1	0	0
1	1	1

A + B *(reads "A or B")* is 1 iff either A or B is 1



A	B	$A + B$
0	0	0
0	1	1
1	0	1
1	1	1

Boolean Algebra: Big Picture



- An algebra on 1's and 0's
  - with AND, OR, NOT operations
- What you start with
  - Axioms: basic things about objects and operations you just assume to be true at the start
- What you derive first
  - Laws and theorems: allow you to manipulate Boolean expressions
  - ...also allow us to do some simplification on Boolean expressions
- What you derive later
  - More "sophisticated" properties useful for manipulating digital designs represented in the form of Boolean equations

George Boole, "The Mathematical Analysis of Logic," 1847.

Boolean Algebra: Axioms

Formal version	*English version*
1. *B* contains at least two elements, *0* and *1*, *such that* $0 \neq 1$	Math formality...
2. *Closure* $a,b \in B$, (i) $a + b \in B$ (ii) $a \bullet b \in B$	Result of AND, OR stays in set you start with
3. *Commutative Laws*: $a,b \in B$, (i) (ii)	For primitive AND, OR of 2 inputs, order doesn't matter
4. *Identities*: $0, 1 \in B$ (i) (ii)	There are identity elements for AND, OR, that give you back what you started with
5. *Distributive Laws*: (i) (ii)	• distributes over +, just like algebra …but + distributes over •, also (!!)
6. *Complement*: (i) (ii)	There is a complement element; AND/ORing with it gives the identity elm.

Boolean Algebra: Duality

- Observation
  - All the axioms come in "dual" form
  - Anything true for an expression also true for its dual
  - So any derivation you could make that is true, can be flipped into dual form, and it stays true
- Duality — More formally
  - A dual of a Boolean expression is derived by replacing
    - Every AND operation with... an OR operation
    - Every OR operation with... an AND
    - Every constant 1 with... a constant 0
    - Every constant 0 with... a constant 1
    - But don't change any of the literals or play with the complements!

Example $a \bullet (b + c) = (a \bullet b) + (a \bullet c)$

$\rightarrow$ $a + (b \bullet c) = (a + b) \bullet (a + c)$

Boolean Algebra: Useful Laws

Dual

Operations with 0 and 1:

1. $X + 0 = X$ — 1D. $X \bullet 1 = X$
2. $X + 1 = 1$ — 2D. $X \bullet 0 = 0$

AND, OR with identities gives you back the original variable or the identity

Idempotent Law:

3. $X + X = X$ — 3D. $X \bullet X = X$

AND, OR with self = self

Involution Law:

4. $\overline{(\overline{X})} = X$

double complement = no complement

Laws of Complementarity:

5. $X + \overline{X} = 1$ — 5D. $X \bullet \overline{X} = 0$

AND, OR with complement gives you an identity

Commutative Law:

6. $X + Y = Y + X$ — 6D. $X \bullet Y = Y \bullet X$

Just an axiom…

Useful Laws (cont)

Associative Laws:

7. $(X + Y) + Z = X + (Y + Z)$
$= X + Y + Z$

7D. $(X \bullet Y) \bullet Z = X \bullet (Y \bullet Z)$
$= X \bullet Y \bullet Z$

Parenthesis order does not matter

Distributive Laws:

8. $X \bullet (Y + Z) = (X \bullet Y) + (X \bullet Z)$

8D. $X + (Y \bullet Z) = (X + Y) \bullet (X + Z)$

Axiom

Simplification Theorems:

9.

9D.

10.

10D.

11.

11D.

Useful for simplifying expressions

Actually worth remembering — they show up a lot in real designs…

Boolean Algebra: Proving Things

Proving theorems via axioms of Boolean Algebra:

EX: Prove the theorem: $X \bullet Y + X \bullet \overline{Y} = X$

Distributive (5)

Complement (6)

Identity (4)

EX2: Prove the theorem: $X + X \bullet Y = X$

Identity (4)

Distributive (5)

Identity (2)

Identity (4)

DeMorgan's Law: Enabling Transformations

DeMorgan's Law:

12. $\overline{(X + Y + Z + \cdots)} = \overline{X}.\overline{Y}.\overline{Z}\ldots$

12D. $\overline{(X.Y.Z\ldots)} = \overline{X} + \overline{Y} + \overline{Z} + \ldots$

- **Think of this as a transformation**
  - Let's say we have:

$$F = A + B + C$$

  - Applying DeMorgan's Law (12), gives us

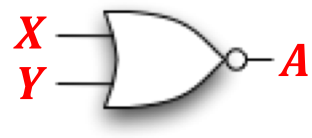
$$F = \overline{\overline{(A + B + C)}} = \overline{(\overline{A}.\overline{B}.\overline{C})}$$

At least one of A, B, C is TRUE --> It is **not** the case that A, B, C are **all** false

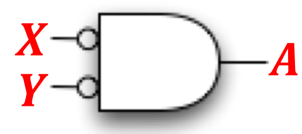
DeMorgan's Law (Continued)

These are conversions between different types of logic functions
They can prove useful if you do not have every type of gate

$$A = \overline{(X+Y)} = \bar{X}\bar{Y}$$

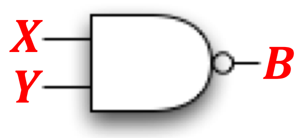


NOR is equivalent to AND with inputs complemented

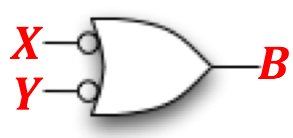


X	Y	$\overline{X+Y}$	$\bar{X}$	$\bar{Y}$	$\bar{X}\bar{Y}$
0	0	1	1	1	1
0	1	0	1	0	0
1	0	0	0	1	0
1	1	0	0	0	0

$$B = \overline{(XY)} = \bar{X} + \bar{Y}$$



NAND is equivalent to OR with inputs complemented



X	Y	$\overline{XY}$	$\bar{X}$	$\bar{Y}$	$\bar{X}+\bar{Y}$
0	0	1	1	1	1
0	1	1	1	0	1
1	0	1	0	1	1
1	1	0	0	0	0

Using Boolean Equations to Represent a Logic Circuit

Sum of Products Form: Key Idea

- Assume we have the truth table of a Boolean Function
- How do we express the function in terms of the inputs in a **standard** manner?
- Idea: **Sum of Products** form
- Express the truth table as a two-level Boolean expression
  - that contains **all** input variable combinations that result in a 1 output
  - If ANY of the combinations of input variables that results in a 1 is TRUE, then the output is 1
  - F = OR of all input variable combinations that result in a 1

Some Definitions

- **Complement:** variable with a bar over it
  $\overline{A}\ ,\ \overline{B}\ ,\ \overline{C}$
- **Literal:** variable or its complement
  $A\ ,\ \overline{A}\ ,\ B\ ,\ \overline{B}\ ,\ C\ ,\ \overline{C}$
- **Implicant:** product (AND) of literals
  $(A \cdot B \cdot \overline{C})\ ,\ (\overline{A} \cdot C)\ ,\ (B \cdot \overline{C})$
- **Minterm:** product (AND) that includes **all** input variables
  $(A \cdot B \cdot \overline{C})\ ,\ (\overline{A} \cdot \overline{B} \cdot C)\ ,\ (\overline{A} \cdot B \cdot \overline{C})$
- **Maxterm:** sum (OR) that includes **all** input variables
  $(A + \overline{B} + \overline{C})\ ,\ (\overline{A} + B + \overline{C})\ ,\ (A + B + \overline{C})$

Two-Level Canonical (Standard) Forms

- Truth table is the unique signature of a Boolean *function* …
  - But, it is an expensive representation

- A Boolean function can have many alternative Boolean expressions
  - i.e., many alternative Boolean expressions (and gate realizations) may have the same truth table (and function)
  - If they all say the same thing, why do we care?
    - Different Boolean expressions lead to different gate realizations

- Canonical form: standard form for a Boolean expression
  - Provides a unique algebraic signature

Two-Level Canonical Forms

Sum of Products Form (SOP)

Also known as **disjunctive normal form** or **minterm expansion**

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

$$F = \overset{0\ 1\ 1}{\underline{\bar{A}BC}} + \overset{1\ 0\ 0}{\underline{A\bar{B}\bar{C}}} + \overset{1\ 0\ 1}{\underline{A\bar{B}C}} + \overset{1\ 1\ 0}{\underline{AB\bar{C}}} + \overset{1\ 1\ 1}{\underline{ABC}}$$

- Each row in a truth table has a minterm
- A minterm is a product (AND) of literals
- Each minterm is TRUE for that row (and only that row)

All Boolean equations can be written in SOP form

Find all the input combinations (minterms) for which the output of the function is TRUE.

SOP Form — Why Does It Work?

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

This input

$$F = \overset{0\ 1\ 1}{\overline{A}BC} + \overset{1\ 0\ 0}{A\overline{B}\,\overline{C}} + \boxed{\overset{1\ 0\ 1}{A\overline{B}C}} + \overset{1\ 1\ 0}{AB\overline{C}} + \overset{1\ 1\ 1}{ABC}$$

Activates this term

- Only the shaded product term — $A\overline{B}C = \mathbf{1 \cdot \overline{0} \cdot 1}$ — will be 1
- No other product terms will "turn on" — they will all be 0
- So if inputs A B C correspond to a product term in expression,
  - We get 0 + 0 + … + 1 + … + 0 + 0 = 1 for output
- If inputs A B C do not correspond to any product term in expression
  - We get 0 + 0 + … + 0 = 0 for output

Aside: Notation for SOP

- Standard "shorthand" notation
  - If we agree on the order of the variables in the rows of truth table…
    - then we can enumerate each row with the decimal number that corresponds to the binary number created by the input pattern

A	B	C	F	
0	0	0	0	
0	0	1	0	
0	1	0	0	
0	1	1	1	
1	0	0	1	**100 = decimal 4 so this is minterm #4, or m4**
1	0	1	1	
1	1	0	1	
1	1	1	1	**111 = decimal 7 so this is minterm #7, or m7**

f =

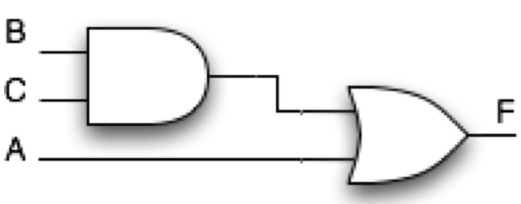
We can write this as a sum of products

Or, we can use a summation notation

Canonical SOP Forms

A	B	C	minterms
0	0	0	$\bar{A}\bar{B}\bar{C}$ = m0
0	0	1	$\bar{A}\bar{B}C$ = m1
0	1	0	$\bar{A}B\bar{C}$ = m2
0	1	1	$\bar{A}BC$ = m3
1	0	0	$A\bar{B}\bar{C}$ = m4
1	0	1	$A\bar{B}C$ = m5
1	1	0	$AB\bar{C}$ = m6
1	1	1	ABC = m7

Shorthand Notation for Minterms of 3 Variables



2-Level AND/OR Realization

F in canonical form:

F(A,B,C) = $\sum$m(3,4,5,6,7)
= m3 + m4 + m5 + m6 + m7

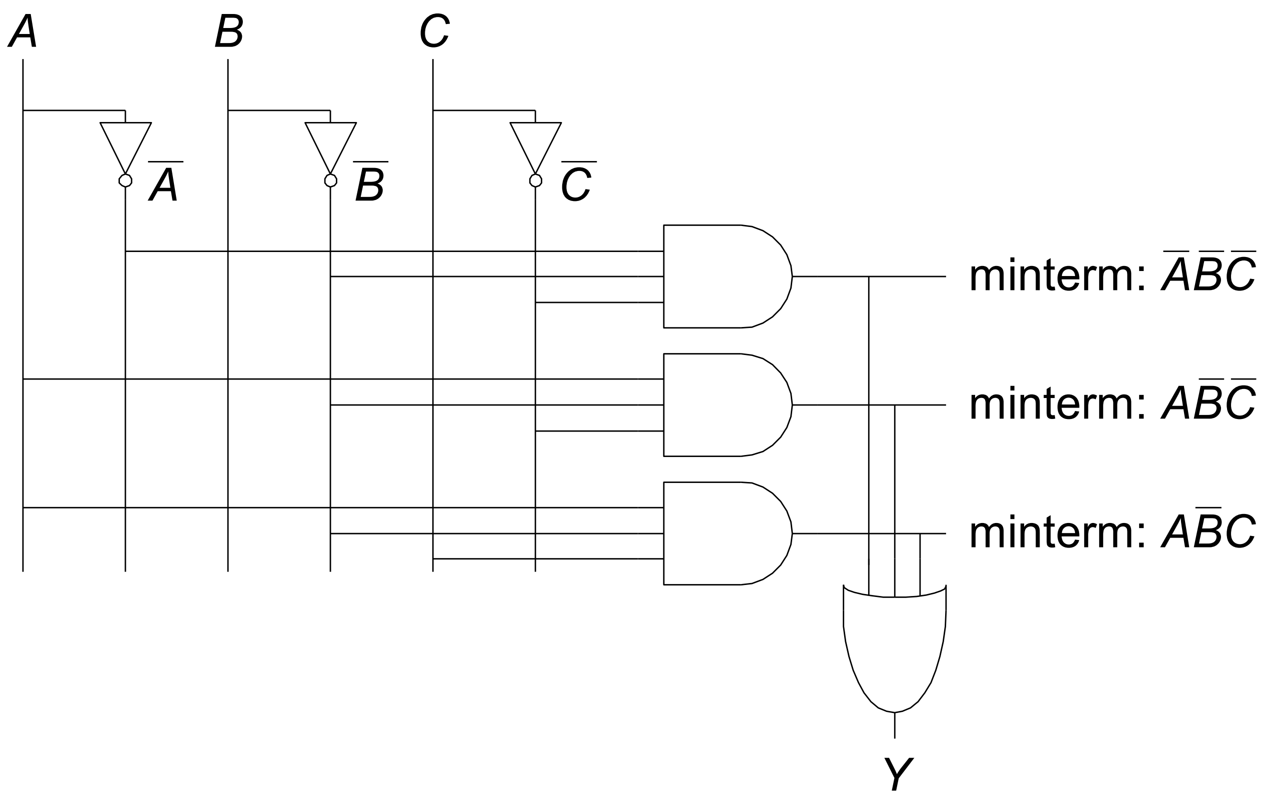
***F* =**

canonical form ≠ minimal form

F

From Logic to Gates

- **SOP (sum-of-products) leads to two-level logic**
- Example: $\boldsymbol{Y = (\overline{A} \cdot \overline{B} \cdot \overline{C}) + (A \cdot \overline{B} \cdot \overline{C}) + (A \cdot \overline{B} \cdot C)}$



Alternative Canonical Form: POS

We can have another from of representation

DeMorgan of SOP of $\overline{F}$

A product of sums (POS)

products

$$F = (A + B + C)(A + B + \overline{C})(A + \overline{B} + C)$$

sums

Each sum term represents one of the "zeros" of the function

This input

$$F = (\overset{0}{A} + \overset{0}{B} + \overset{0}{C})\ (\overset{0}{A} + \overset{0}{B} + \overset{1}{\overline{C}})\ (\overset{0}{A} + \overset{1}{\overline{B}} + \overset{0}{C})$$

Activates this term

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

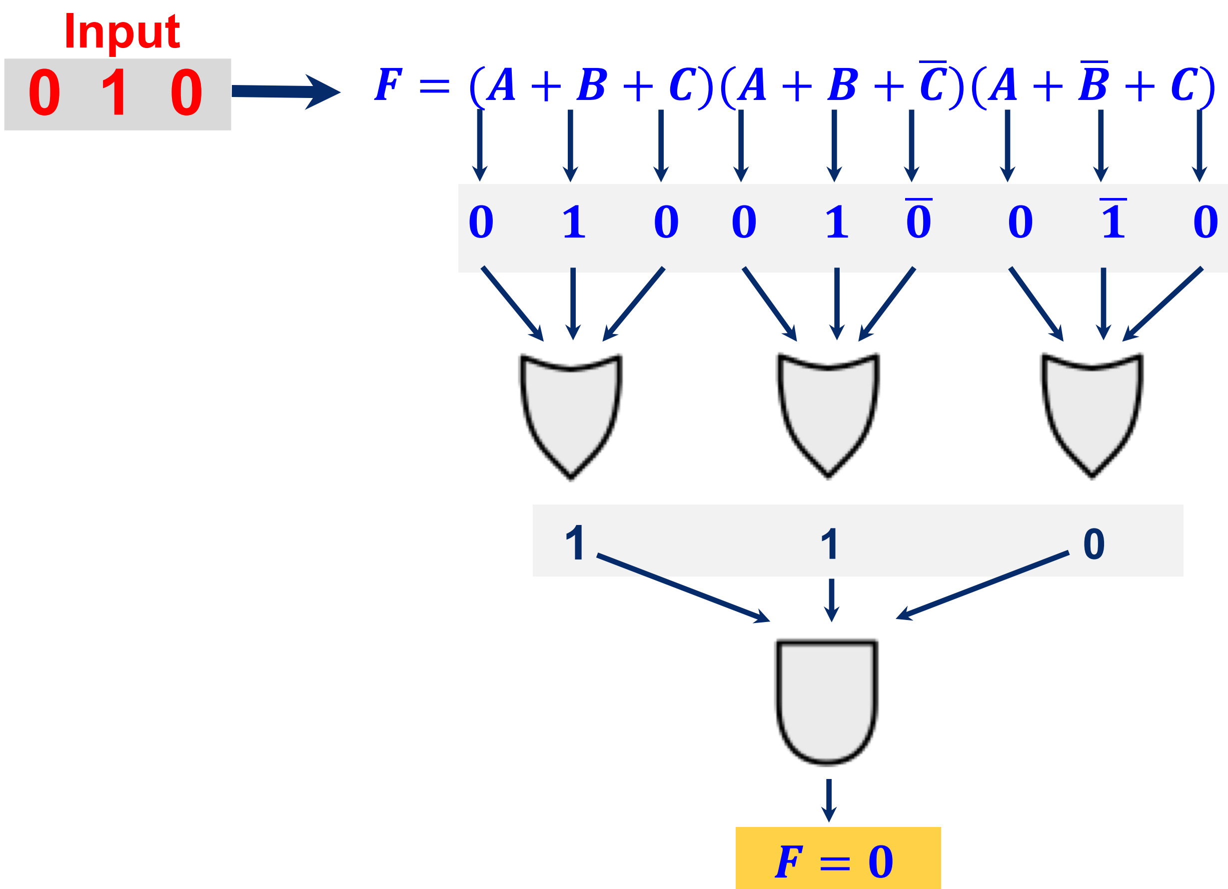
For the given input, only the shaded sum term will equal 0

$$A + \overline{B} + C = 0 + \overline{1} + 0$$

Anything ANDed with 0 is 0; Output F will be 0

Consider A=0, B=1, C=0

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1



Only one of the products will be 0, anything ANDed with 0 is 0

Therefore, the output is F = 0

POS: How to Write It

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

$$F = (A + B + C)(A + B + \overline{C})(A + \overline{B} + C)$$

$A \quad \overline{B} \quad C$

$A + \overline{B} + C$

Maxterm form:

1. Find truth table rows where F is 0

2. 0 in input col → true literal
3. 1 in input col → complemented literal

4. OR the literals to get a Maxterm
5. AND together all the Maxterms

Or just remember, POS of F is the same as the DeMorgan of SOP of $\overline{F}$!!

Canonical POS Forms

Product of Sums / Conjunctive Normal Form / Maxterm Expansion

A	B	C	Maxterms	
0	0	0	$A+B+C$	= M0
0	0	1	$A+B+\overline{C}$	= M1
0	1	0	$A+\overline{B}+C$	= M2
0	1	1	$A+\overline{B}+\overline{C}$	= M3
1	0	0	$\overline{A}+B+C$	= M4
1	0	1	$\overline{A}+B+\overline{C}$	= M5
1	1	0	$\overline{A}+\overline{B}+C$	= M6
1	1	1	$\overline{A}+\overline{B}+\overline{C}$	= M7

Maxterm shorthand notation for a function of three variables

$$\mathbf{F} = (A+B+C)(A+B+\overline{C})(A+\overline{B}+C)$$

$$\prod M(0,1,2)$$

A	B	C	F
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1

Note that you form the maxterms around the "zeros" of the function

This is not the complement of the function!

Useful Conversions

1. **Minterm to Maxterm conversion:**
   rewrite minterm shorthand using maxterm shorthand
   replace minterm indices with the indices not already used
   E.g., $\mathrm{F}(A, B, C) = \sum m(3, 4, 5, 6, 7) = \prod M(0, 1, 2)$

2. **Maxterm to Minterm conversion:**
   rewrite maxterm shorthand using minterm shorthand
   replace maxterm indices with the indices not already used
   E.g., $\mathrm{F}(A, B, C) = \prod M(0, 1, 2) = \sum m(3, 4, 5, 6, 7)$

3. **Expansion of F to expansion of $\bar{F}$:**

$$\mathrm{E.g.,}\ \mathrm{F}(A, B, C) = \sum m(3, 4, 5, 6, 7) \longrightarrow \bar{F}(A, B, C) = \sum m(0, 1, 2)$$

$$= \prod M(0, 1, 2) \longrightarrow = \prod M(3, 4, 5, 6, 7)$$

4. **Minterm expansion of F to Maxterm expansion of $\bar{F}$:**
   rewrite in Maxterm form, using the same indices as F

$$\mathrm{E.g.,}\ \mathrm{F}(A, B, C) = \sum m(3, 4, 5, 6, 7) \longrightarrow \bar{F}(A, B, C) = \prod M(3, 4, 5, 6, 7)$$

$$= \prod M(0, 1, 2) \longrightarrow = \sum m(0, 1, 2)$$

Combinational Building Blocks used in Modern Computers

Combinational Building Blocks

- Combinational logic is often grouped into larger building blocks to build more complex systems

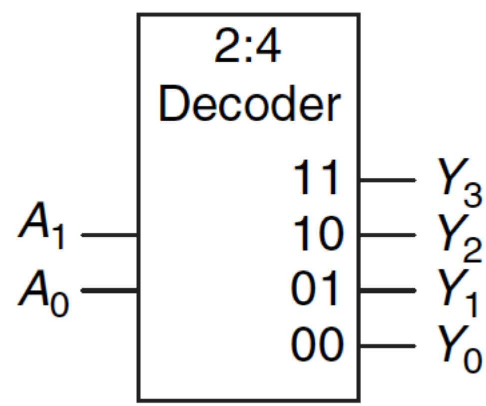
- Hides the unnecessary gate-level details to emphasize the function of the building block

- We now look at:
  - Decoder
  - Multiplexer
  - Full adder
  - PLA (Programmable Logic Array)

Decoder

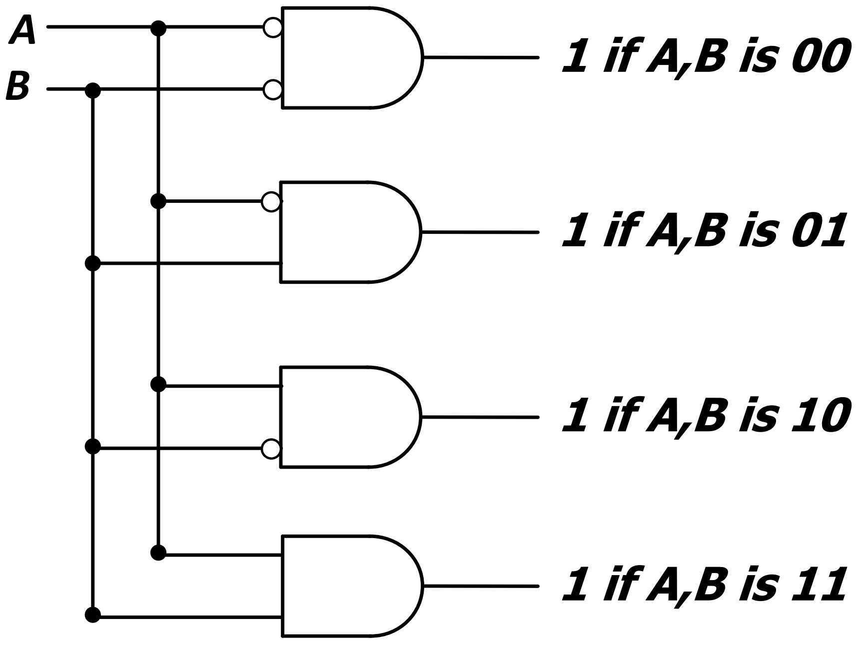
- "Input pattern detector"
- n inputs and 2^n outputs
- Exactly one of the outputs is 1 and all the rest are 0s
- The one output that is logically 1 is the output corresponding to the input pattern that the logic circuit is expected to detect
- Example: 2-to-4 decoder

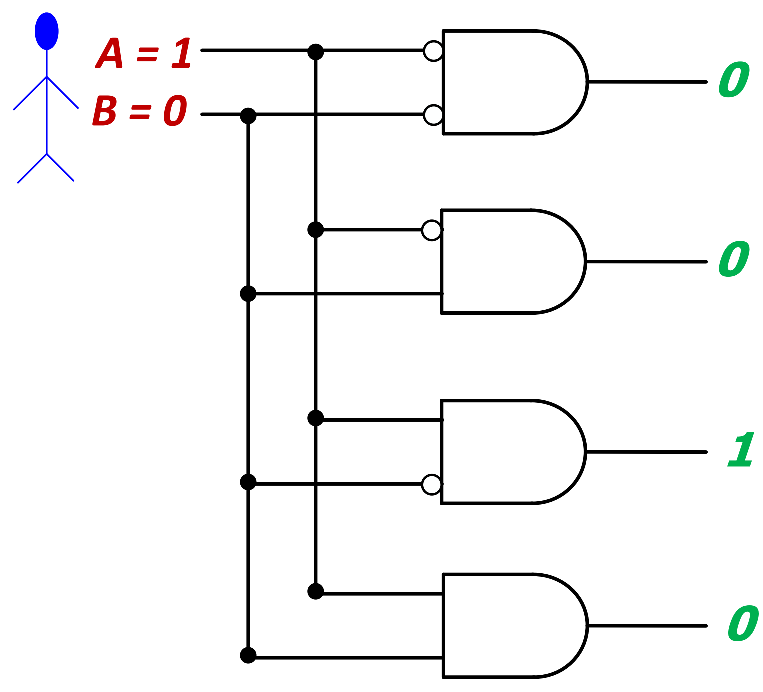
A_1	A_0	Y_3	Y_2	Y_1	Y_0
0	0	0	0	0	1
0	1	0	0	1	0
1	0	0	1	0	0
1	1	1	0	0	0



Decoder (I)

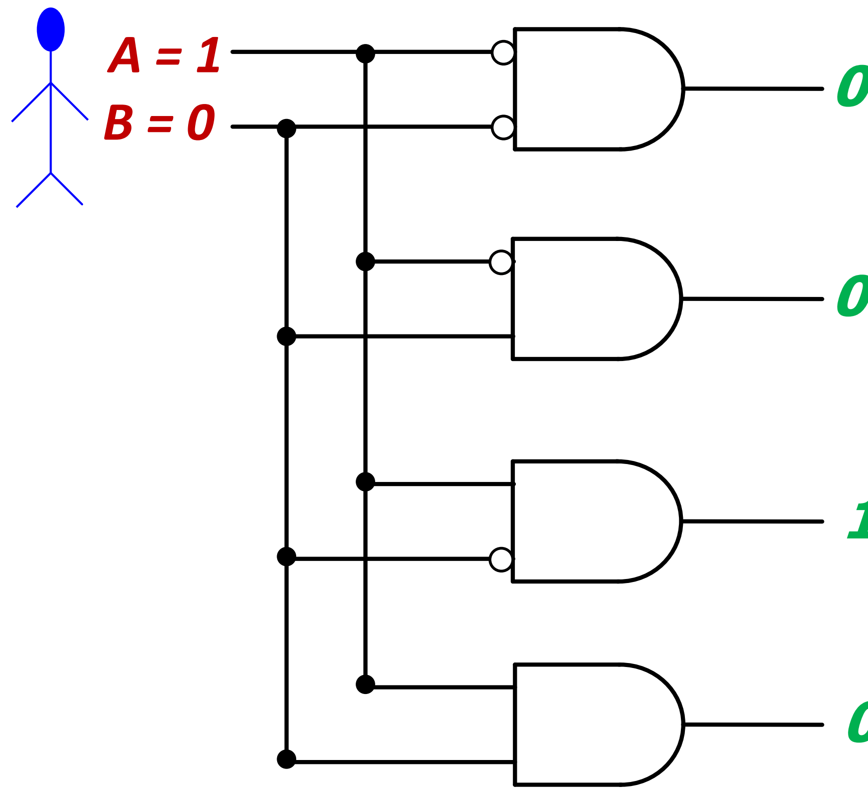
- n inputs and 2^n outputs
- Exactly one of the outputs is 1 and all the rest are 0s
- The one output that is logically 1 is the output corresponding to the input pattern that the logic circuit is expected to detect





Decoder (II)

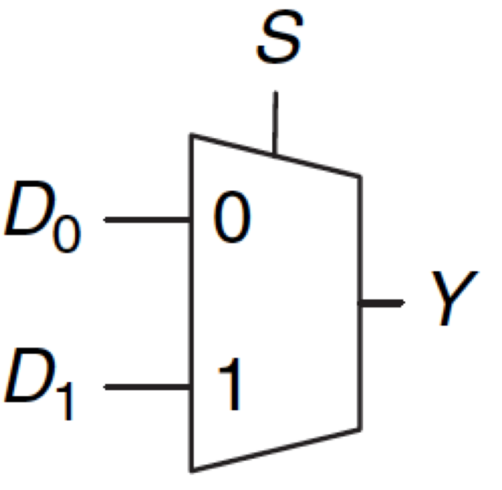
- The decoder is useful in determining how to interpret a bit pattern
  - **It could be the address of a row in DRAM, that the processor intends to read from**
  - **It could be an instruction in the program and the processor has to decide what action to do! (based on *instruction opcode*)**



Multiplexer (MUX), or Selector

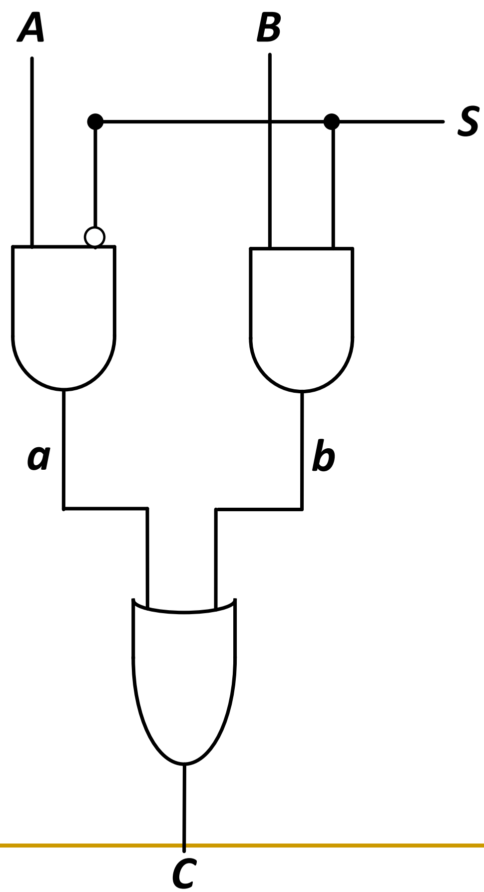
- Selects one of the *N* inputs to connect it to the output
  - based on the value of a $\log_2 N$-bit control input called select
- Example: 2-to-1 MUX

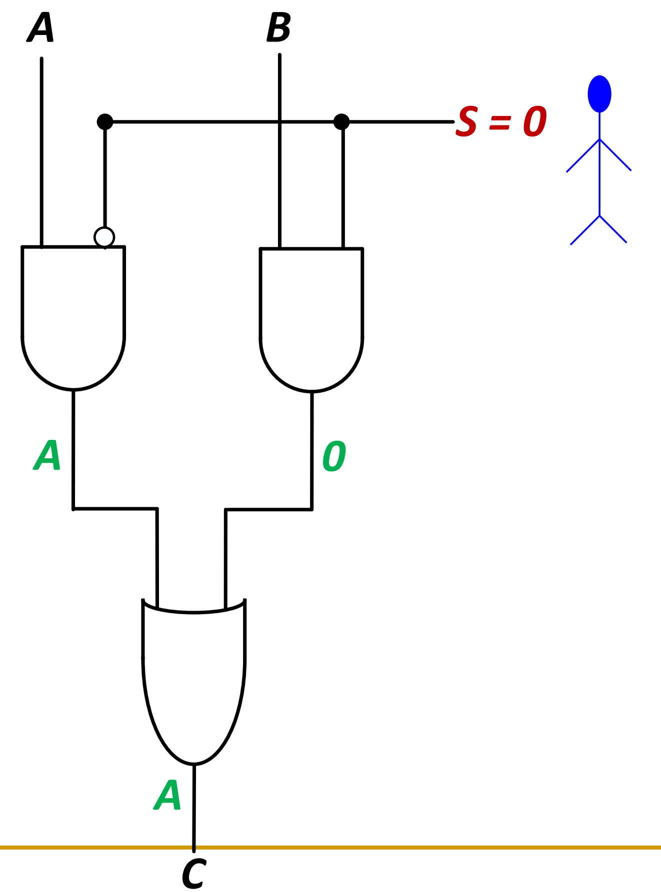
S	D_1	D_0	Y
0	0	0	0
0	0	1	1
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	0
1	1	0	1
1	1	1	1



Multiplexer (MUX), or Selector (II)

- Selects one of the *N* inputs to connect it to the output
  - based on the value of a $\log_2 N$-bit control input called select
- Example: 2-to-1 MUX

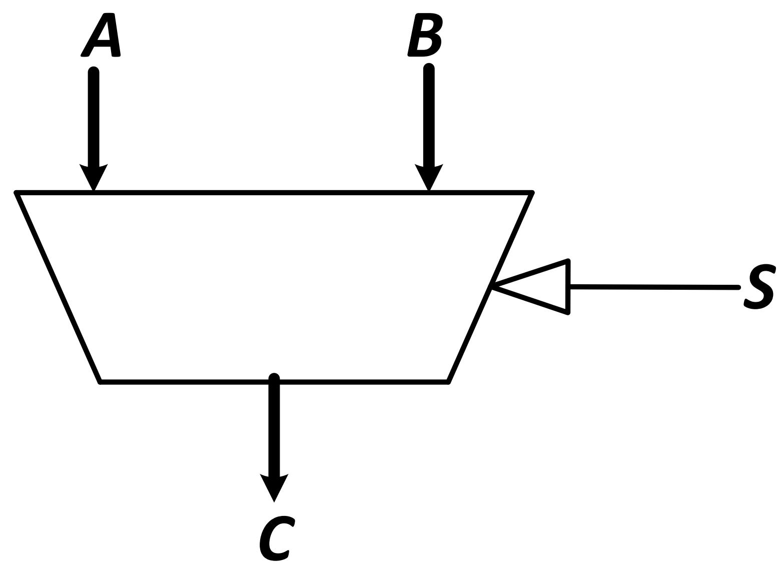




Multiplexer (MUX), or Selector (III)

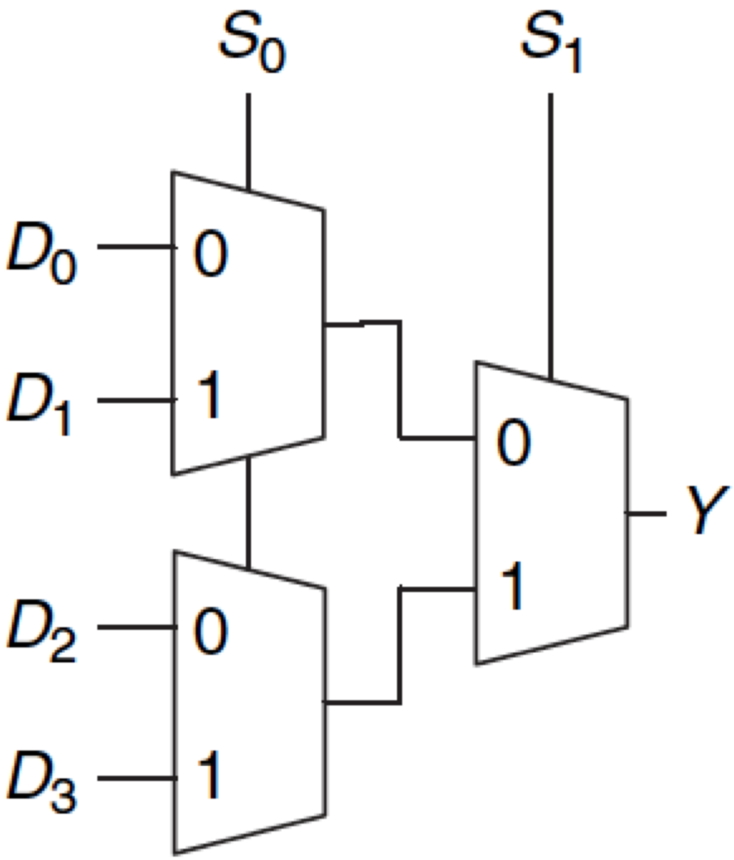
- The output C is always connected to either the input A or the input B
  - Output value depends on the value of the select line S

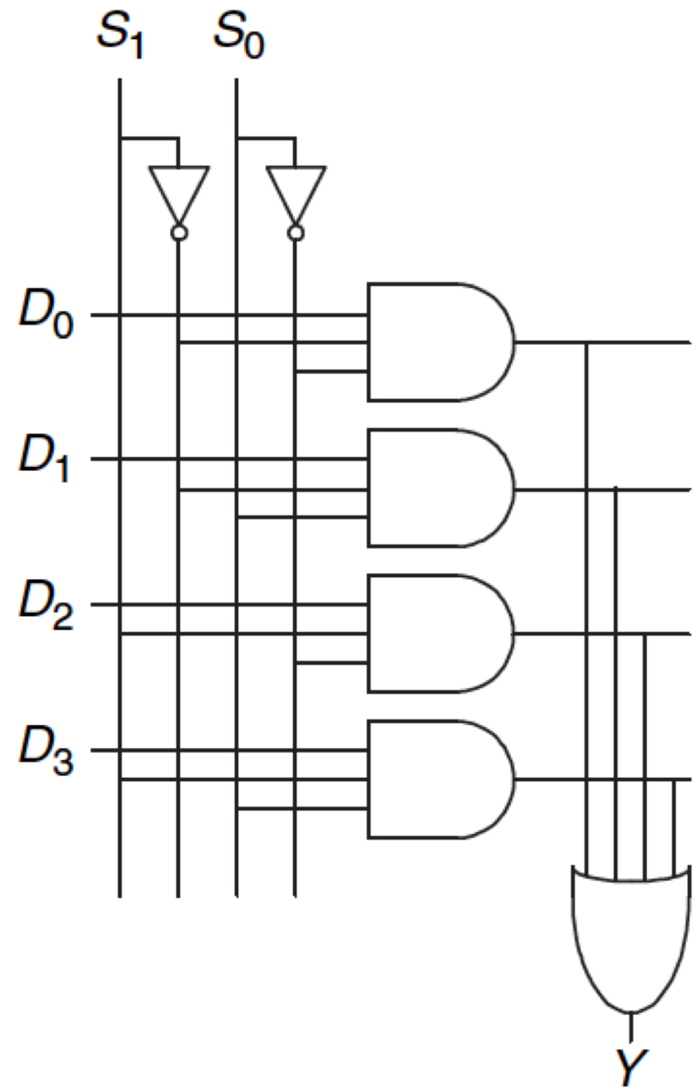
S	C
0	A
1	B



- Your task: Draw the schematic for an 4-input (4:1) MUX
  - Gate level: as a combination of basic AND, OR, NOT gates
  - Module level: As a combination of 2-input (2:1) MUXes

A 4-to-1 Multiplexer





Full Adder (I)

- **Binary addition**
  - Similar to decimal addition
  - From right to left
  - One column at a time
  - One sum and one carry bit

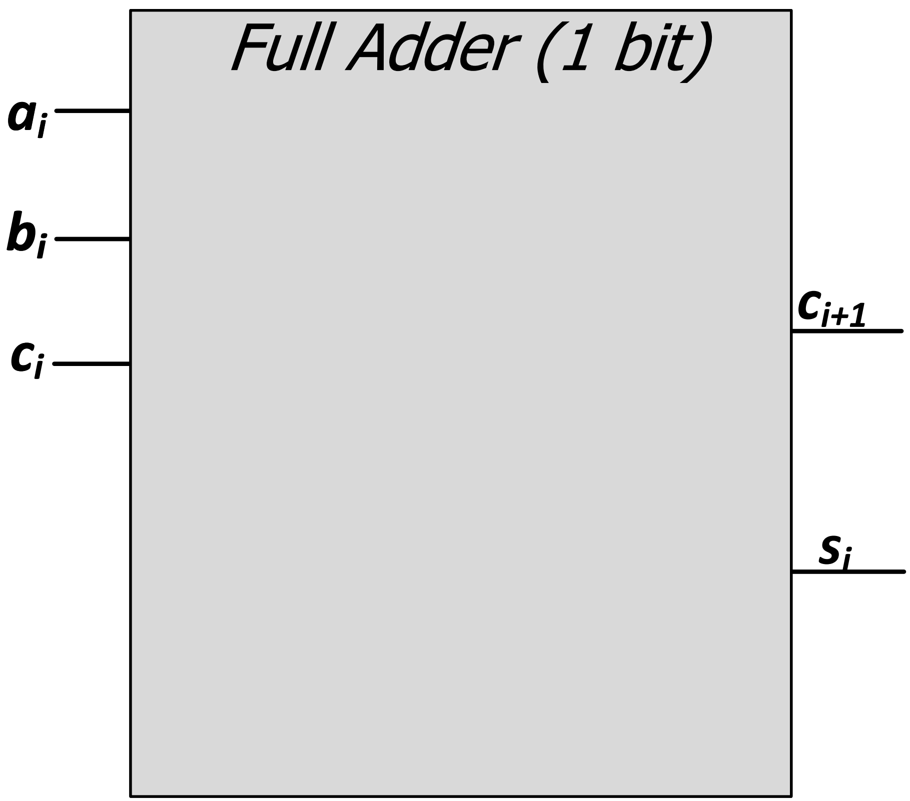
$$
\begin{array}{r}
a_{n-1}a_{n-2}\ldots a_1 a_0 \\
b_{n-1}b_{n-2}\ldots b_1 b_0 \\
C_n\ C_{n-1}\quad \ldots\quad C_1\ \ \\
\hline
S_{n-1}\quad \ldots\quad S_1 S_0
\end{array}
$$

- Truth table of binary addition on one column of bits within two n-bit operands

a_i	b_i	$carry_i$	$carry_{i+1}$	S_i
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

Full Adder (II)

- **Binary addition**
  - N 1-bit additions
  - **SOP of 1-bit addition**

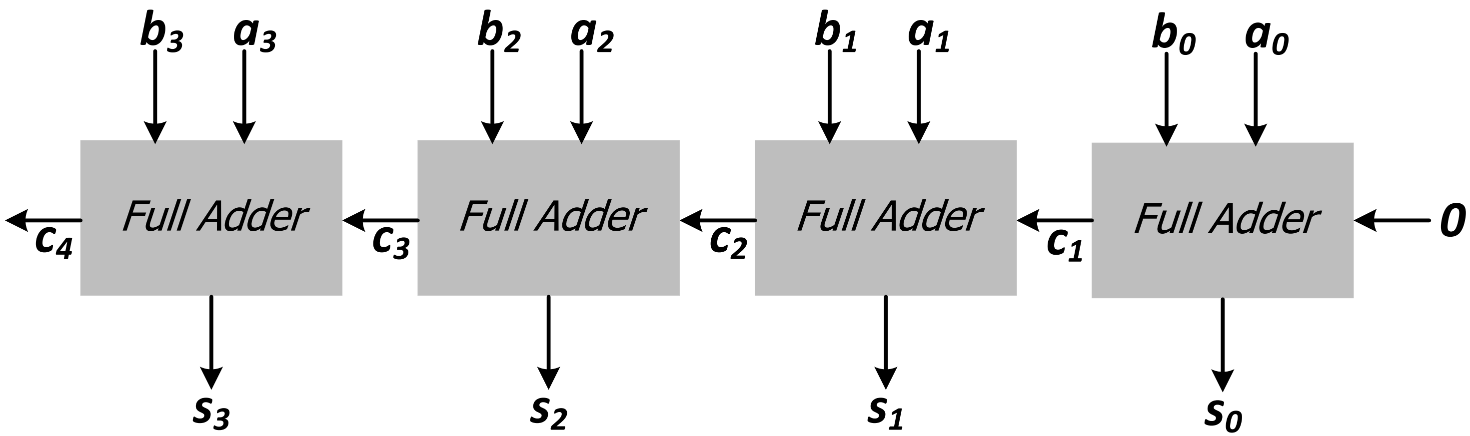


$$
\begin{array}{r}
a_{n-1}a_{n-2}\ldots a_1a_0 \\
b_{n-1}b_{n-2}\ldots b_1b_0 \\
C_n\,C_{n-1}\quad\ldots\quad C_1 \\
\hline
S_{n-1}\quad\ldots\quad S_1S_0
\end{array}
$$

a_i	b_i	$carry_i$	$carry_{i+1}$	S_i
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

4-Bit Adder from Full Adders

- Creating a **4-bit adder** out of 1-bit full adders
  - To add two 4-bit binary numbers A and B



$$
\begin{array}{rrrrr}
 & a_3 & a_2 & a_1 & a_0 \\
+ & b_3 & b_2 & b_1 & b_0 \\
c_4 & c_3 & c_2 & c_1 & \\
\hline
 & s_3 & s_2 & s_1 & s_0
\end{array}
\qquad\qquad
\begin{array}{rrrrr}
 & 1 & 0 & 1 & 1 \\
+ & 1 & 0 & 0 & 1 \\
1 & 0 & 1 & 1 & \\
\hline
 & 0 & 1 & 0 & 0
\end{array}
$$

Adder Design: Ripple Carry Adder

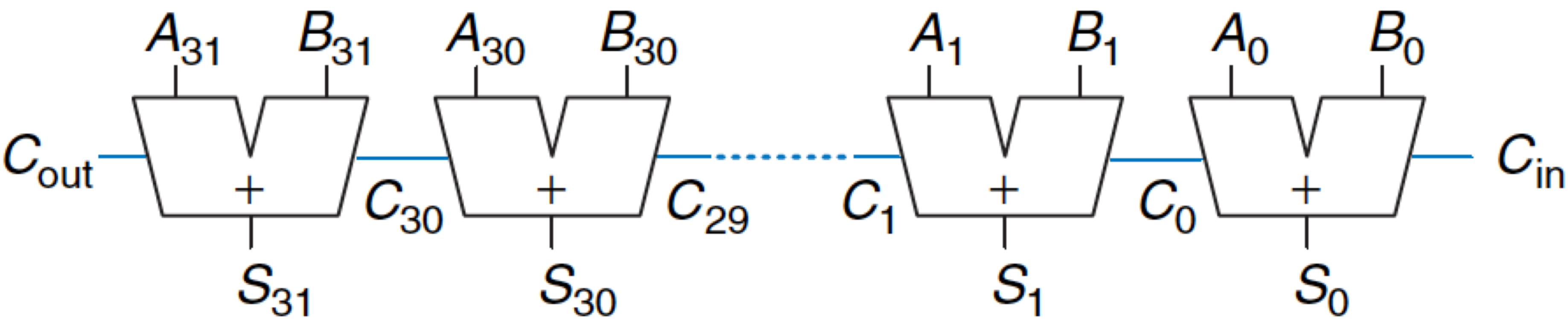
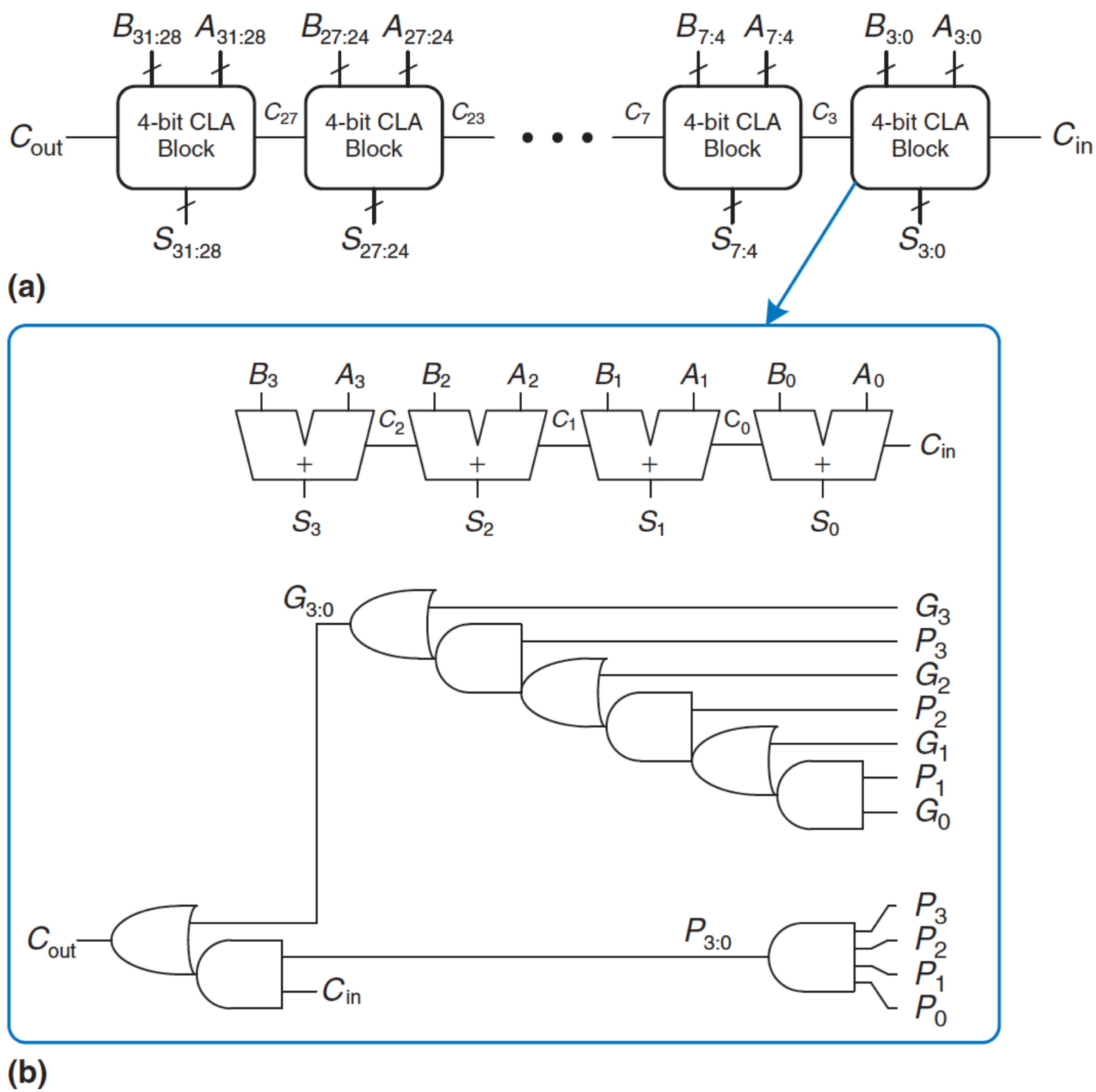


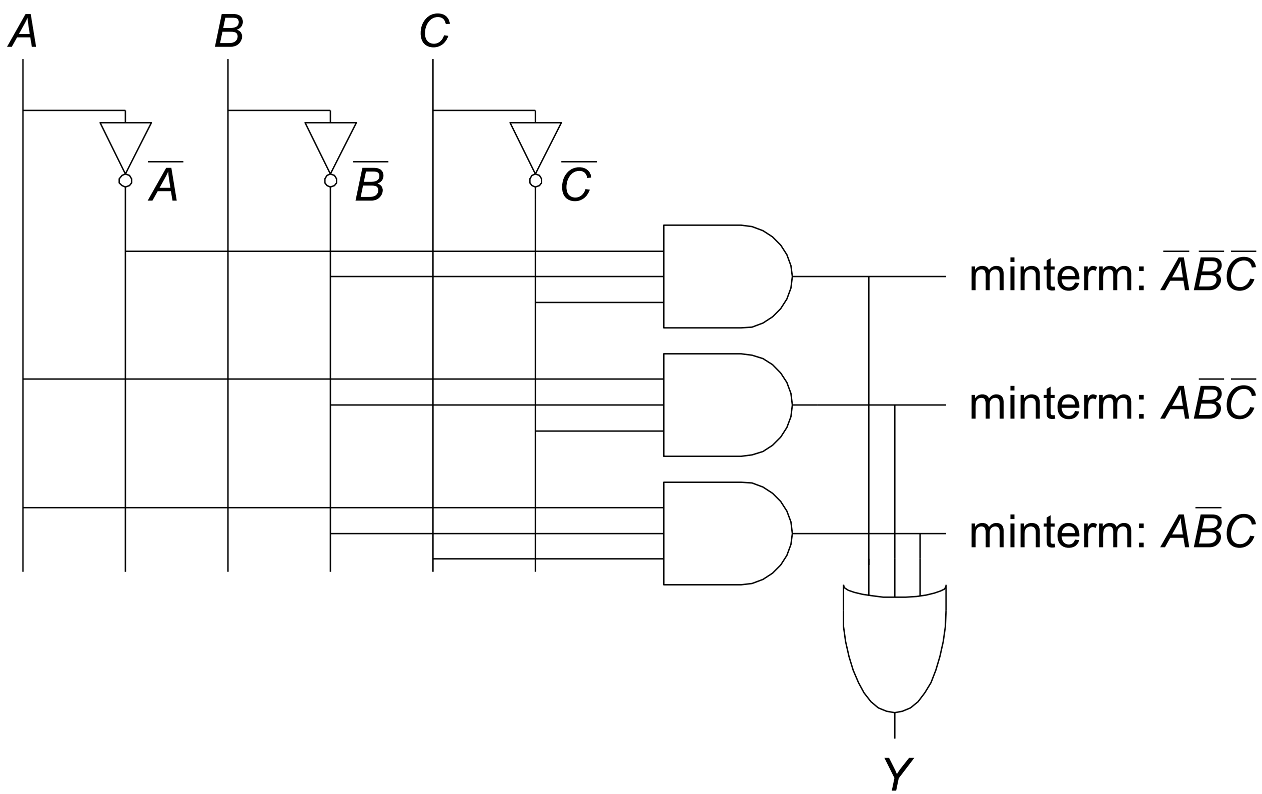
Figure 5.5 **32-bit ripple-carry adder**

Adder Design: Carry Lookahead Adder



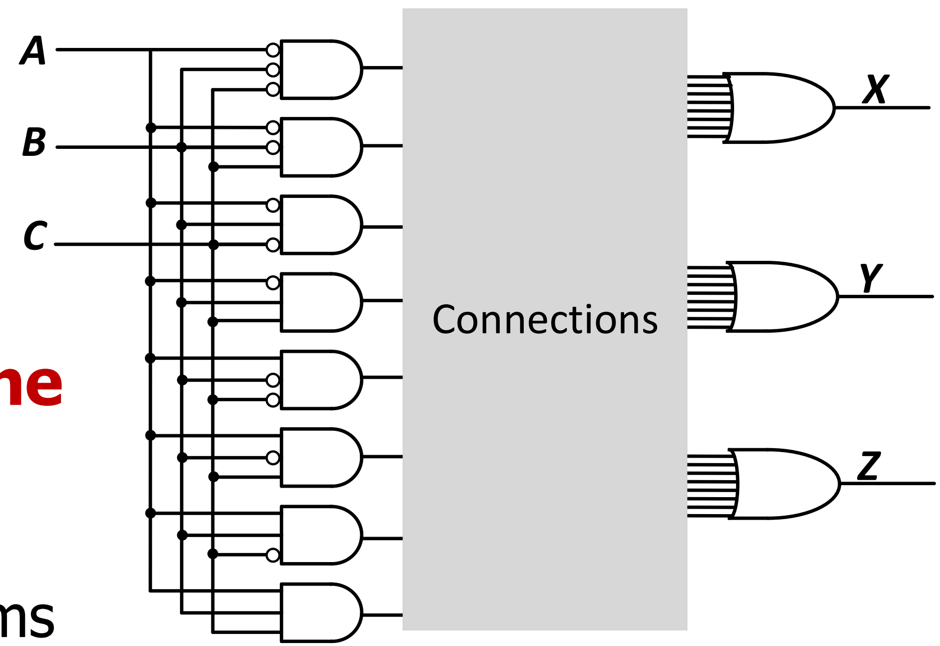
PLA: Recall: From Logic to Gates

- **SOP (sum-of-products) leads to two-level logic**
- Example: $\boldsymbol{Y = (\overline{A} \cdot \overline{B} \cdot \overline{C}) + (A \cdot \overline{B} \cdot \overline{C}) + (A \cdot \overline{B} \cdot C)}$



The Programmable Logic Array (PLA)

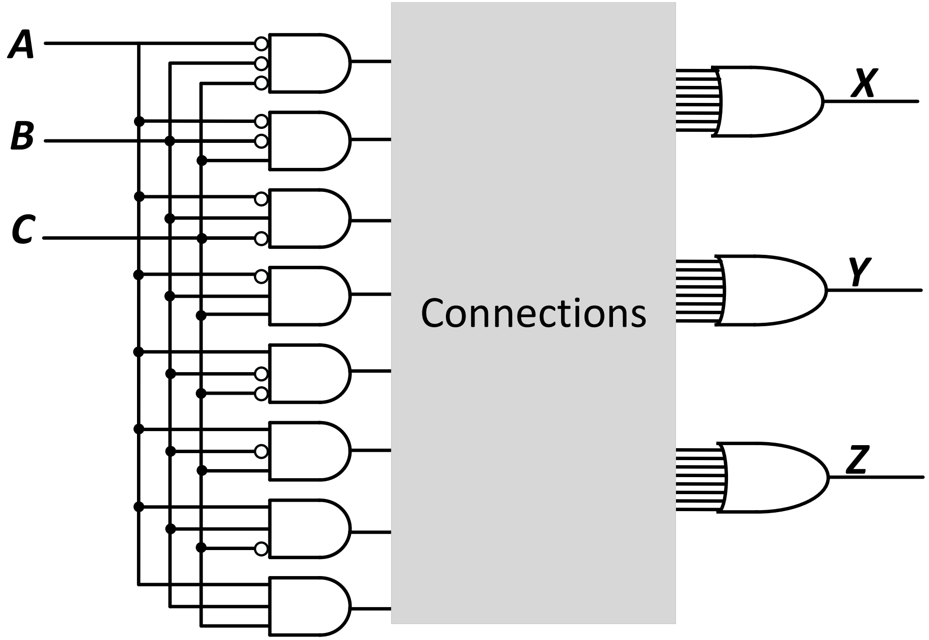
- The below logic structure is a very common building block for implementing any collection of logic functions one wishes to



- An array of AND gates followed by an array of OR gates
- **How do we determine the number of AND gates?**
  - **Remember SOP:** the number of possible minterms
  - For an n-input logic function, we need a PLA with 2^n n-input AND gates
- **How do we determine the number of OR gates?** The number of output columns in the truth table

The Programmable Logic Array (PLA)

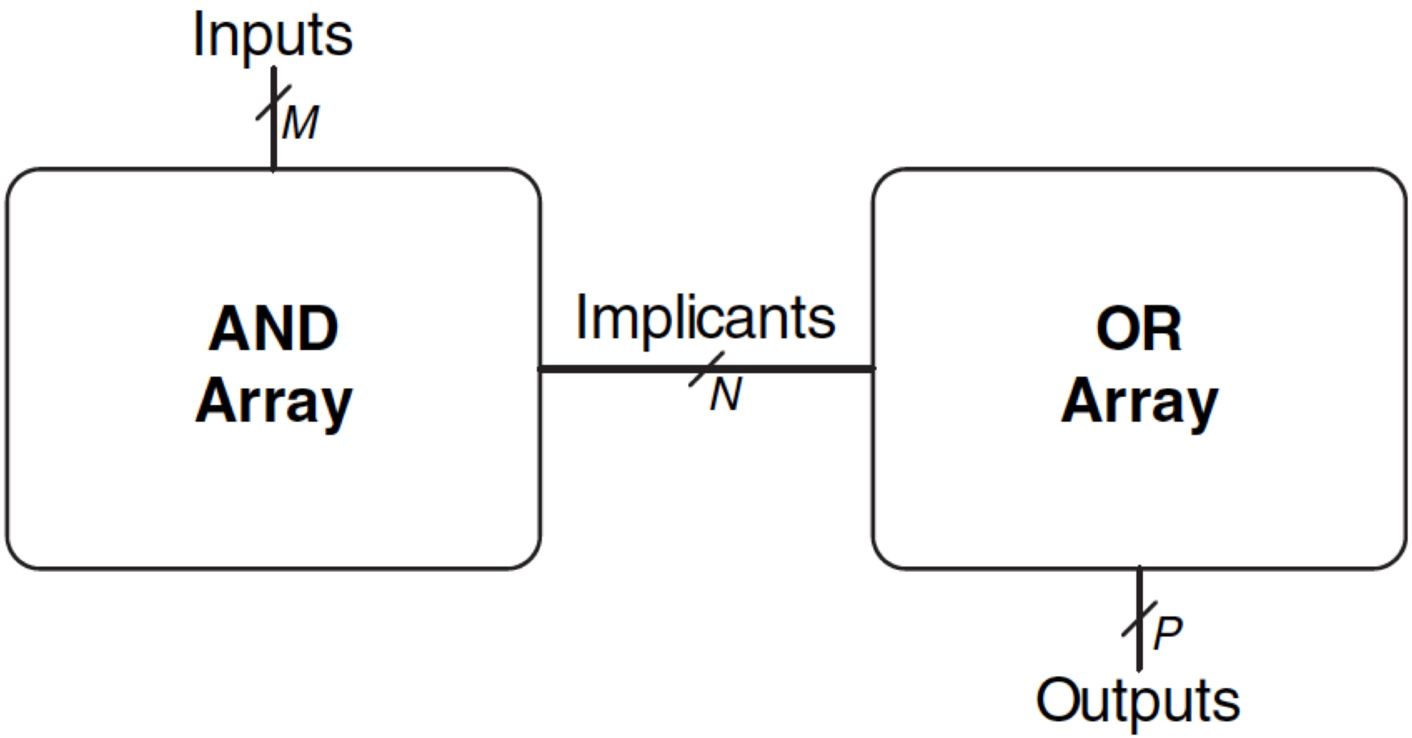
- How do we implement a logic function?
  - Connect the output of an AND gate to the input of an OR gate if the corresponding minterm is included in the SOP
  - This is a simple programmable logic



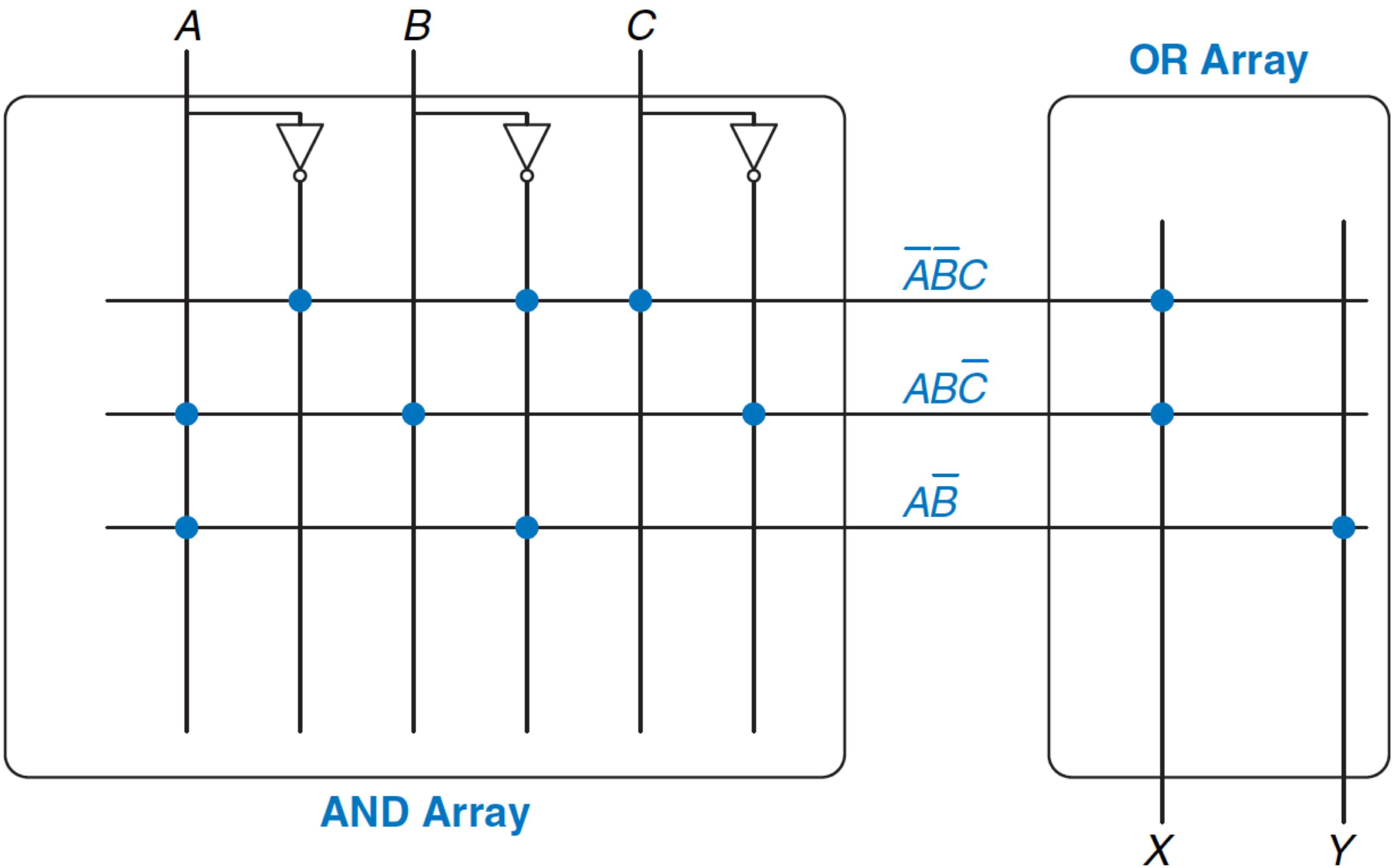
- **Programming a PLA**: we program the connections from AND gate outputs to OR gate inputs to implement a desired logic function

- Have you seen any other type of programmable logic?
  - Yes! An FPGA…
  - An FPGA uses more advanced structures, as we saw in Lecture 3

PLA Example (I)

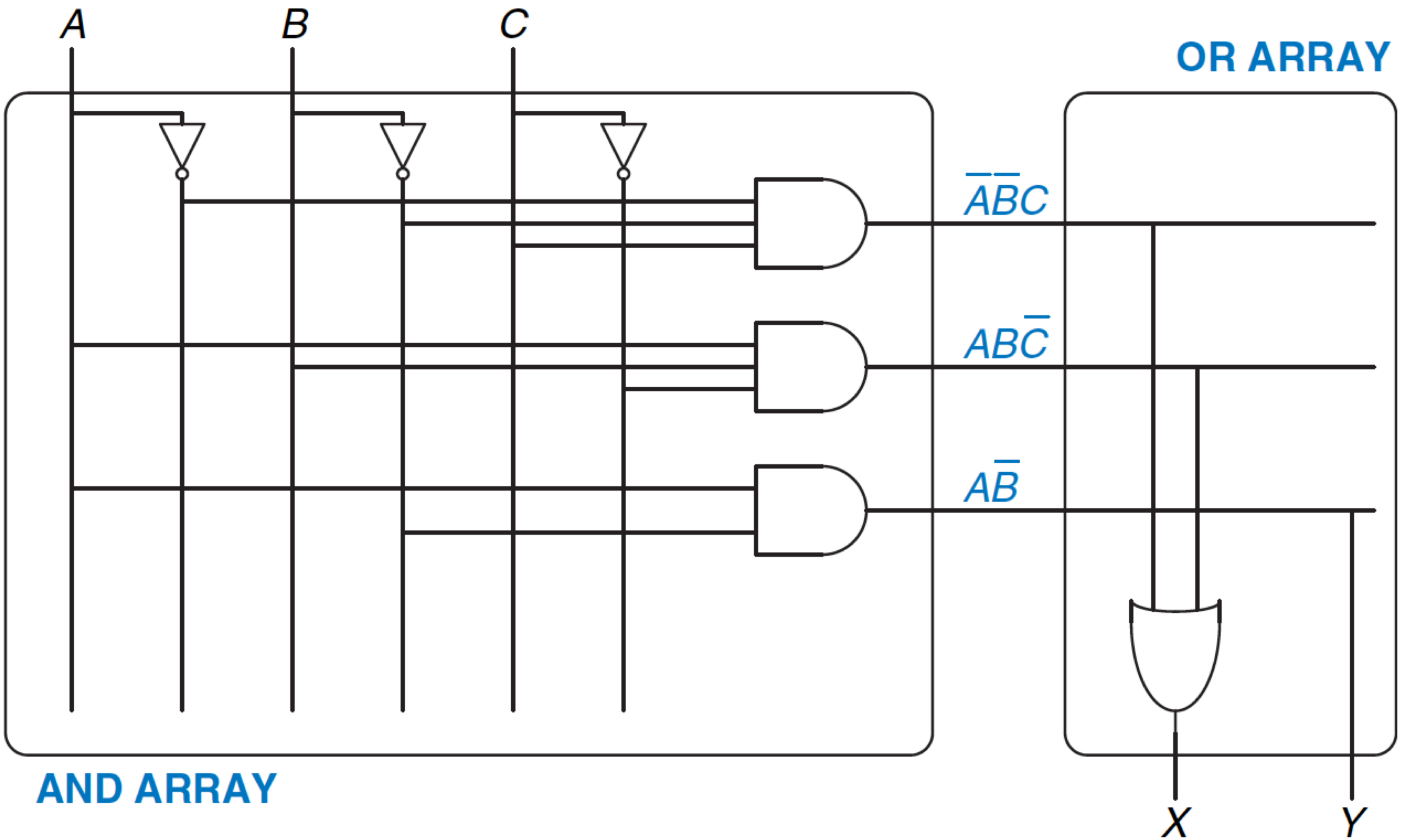


PLA Example Function (II)



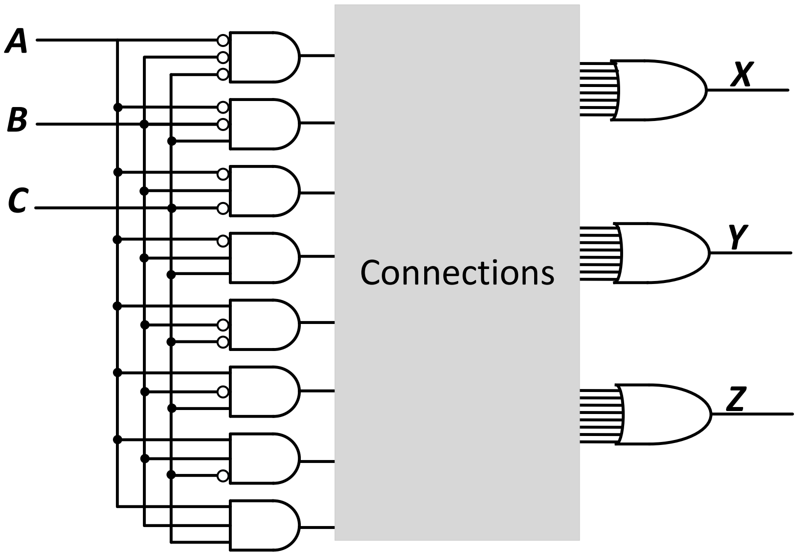
Read H&H Chapter 5.6.1

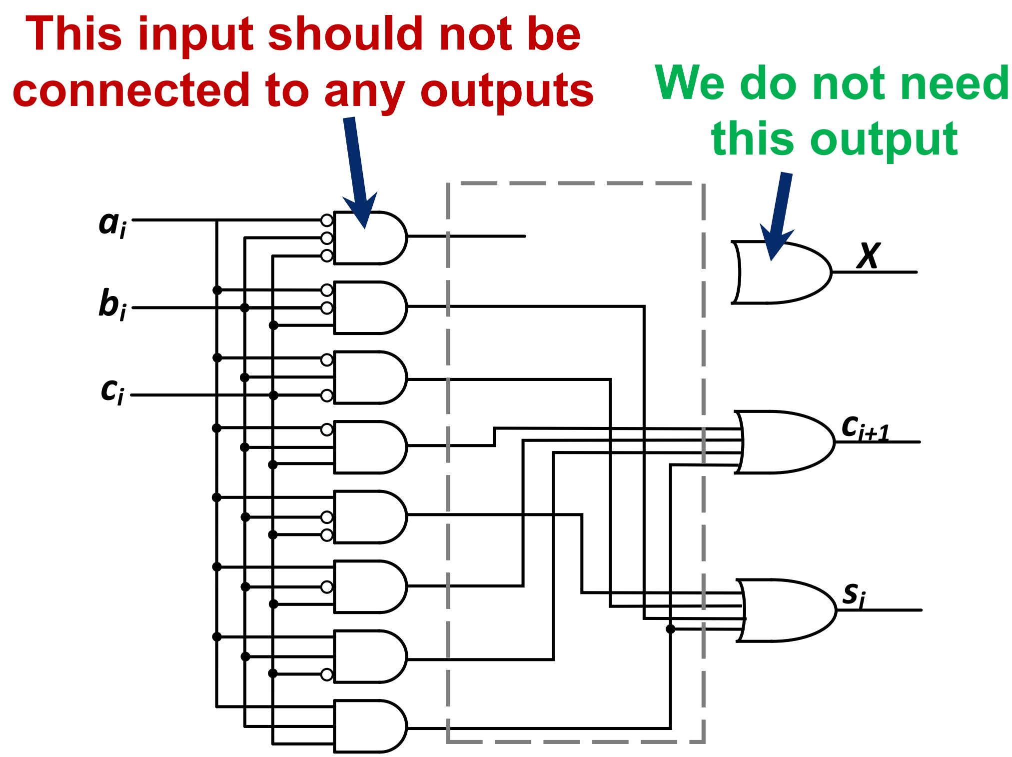
PLA Example Function (III)



Read H&H Chapter 5.6.1

Implementing a Full Adder Using a PLA





Truth table of a full adder

a_i	b_i	$carry_i$	$carry_{i+1}$	S_i
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

Logical (Functional) Completeness

- Any logic function we wish to implement could be accomplished with a PLA
  - PLA consists of only AND gates, OR gates, and inverters
  - We just have to program connections based on SOP of the intended logic function

- The set of gates {AND, OR, NOT} is logically complete because we can build a circuit to carry out the specification of any truth table we wish, without using any other kind of gate

- NAND is also logically complete. So is NOR.
  - Your task: Prove this.

More Combinational Building Blocks

- H&H Chapter 2 in full
  - Required Reading
  - E.g., see Tri-state Buffer and Z values in Section 2.6

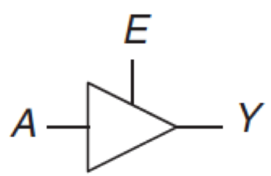
- H&H Chapter 5
  - Will be required reading soon.

- You will benefit greatly by reading the "combinational" parts of Chapter 5 soon.
  - Sections 5.1 and 5.2

Tri-State Buffer

- A tri-state buffer enables gating of different signals onto a wire

Tristate Buffer



E	A	Y
0	0	Z
0	1	Z
1	0	0
1	1	1

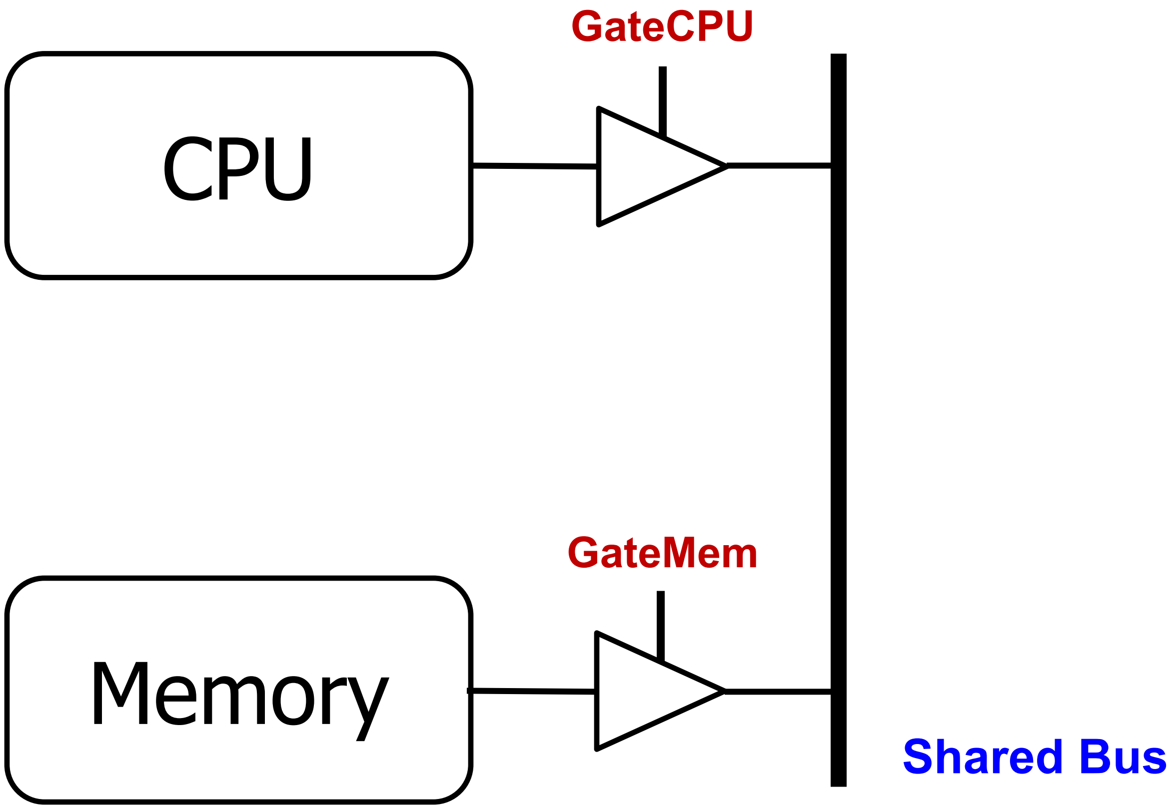
Figure 2.40 Tristate buffer

- Floating signal (Z): Signal that is not driven by any circuit
  - Open circuit, floating wire

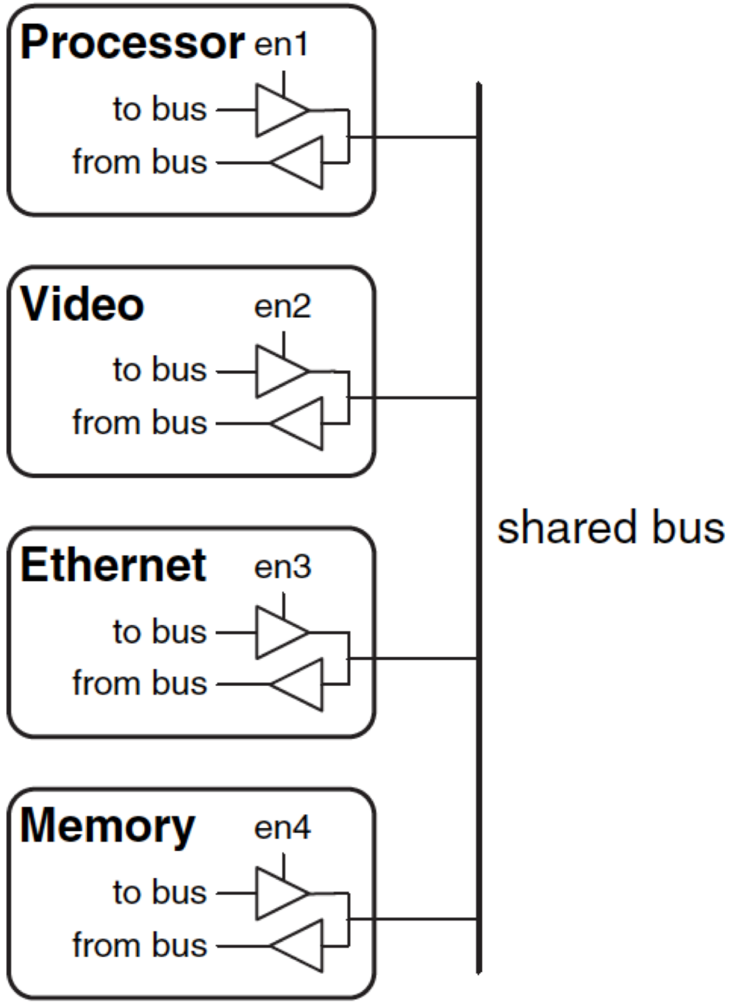
Example: Use of Tri-State Buffers

- Imagine a wire connecting the CPU and memory
  - At any time only the CPU or the memory can place a value on the wire, both not both
  - You can have two tri-state buffers: one driven by CPU, the other memory; and ensure at most one is enabled at any time

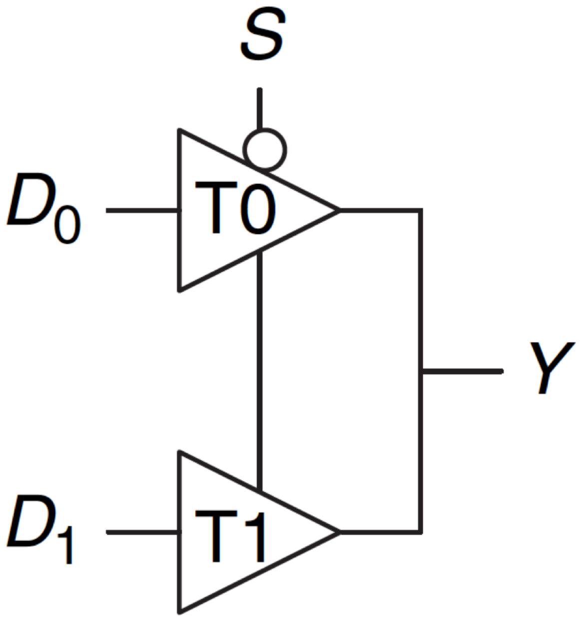
Example Design with Tri-State Buffers



Another Example

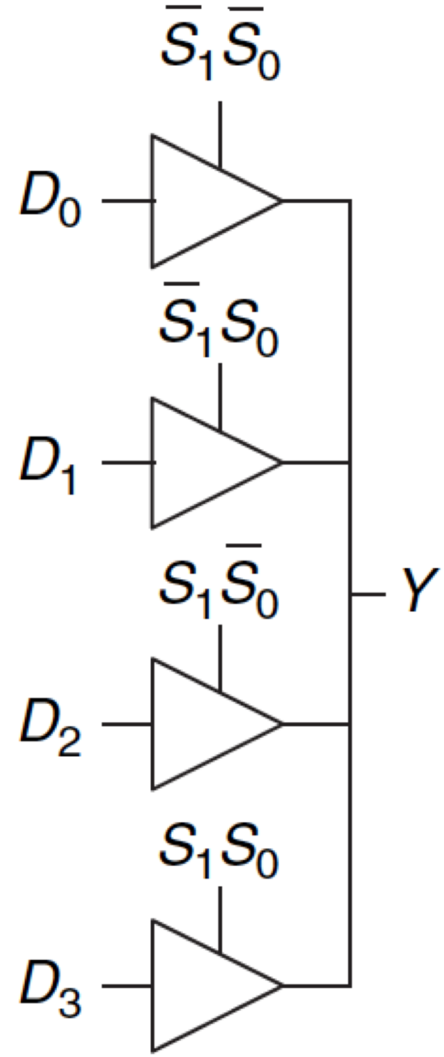


Multiplexer Using Tri-State Buffers



$$Y = D_0\overline{S} + D_1 S$$

Figure 2.56 **Multiplexer using tristate buffers**



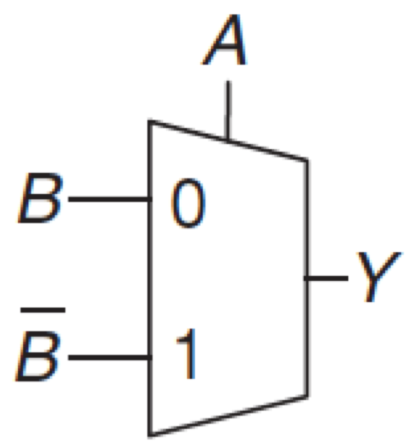
Aside: Logic Using Multiplexers (II)

- Multiplexers can be used as lookup tables to perform logic functions

$$Y = A \oplus B$$

A	B	Y
0	0	0
0	1	1
1	0	1
1	1	0

A	Y
0	B
1	$\overline{B}$

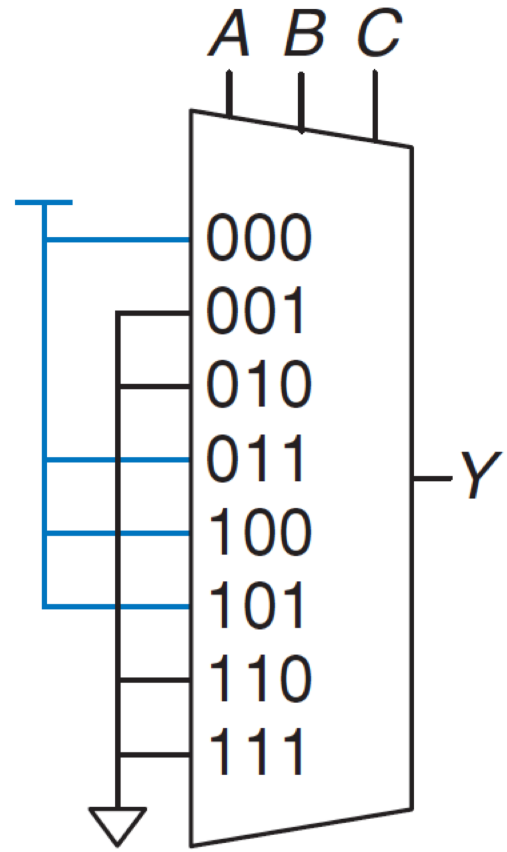


Aside: Logic Using Multiplexers (III)

- Multiplexers can be used as lookup tables to perform logic functions

A	B	C	Y
0	0	0	1
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	0
1	1	1	0

$$Y = A\overline{B} + \overline{B}\,\overline{C} + \overline{A}BC$$



Aside: Logic Using Decoders (I)

- Decoders can be combined with OR gates to build logic functions.

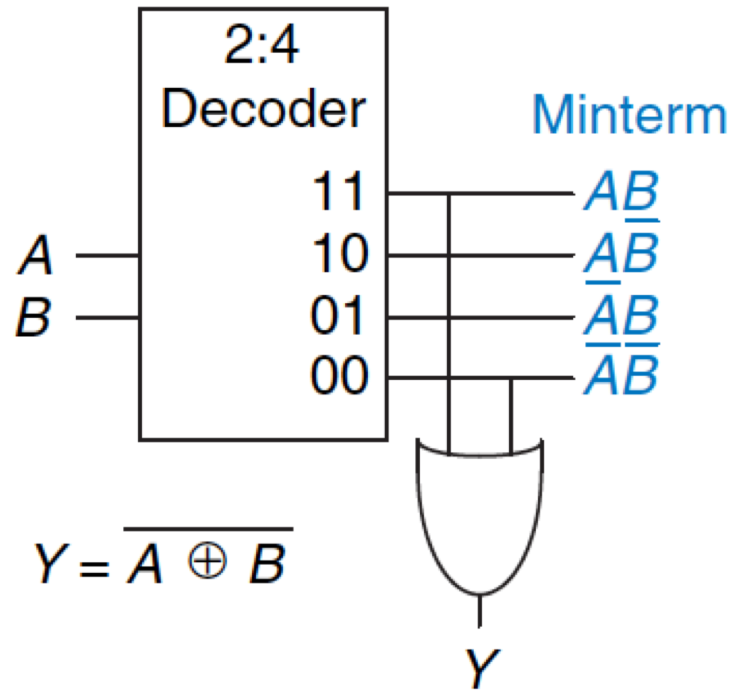
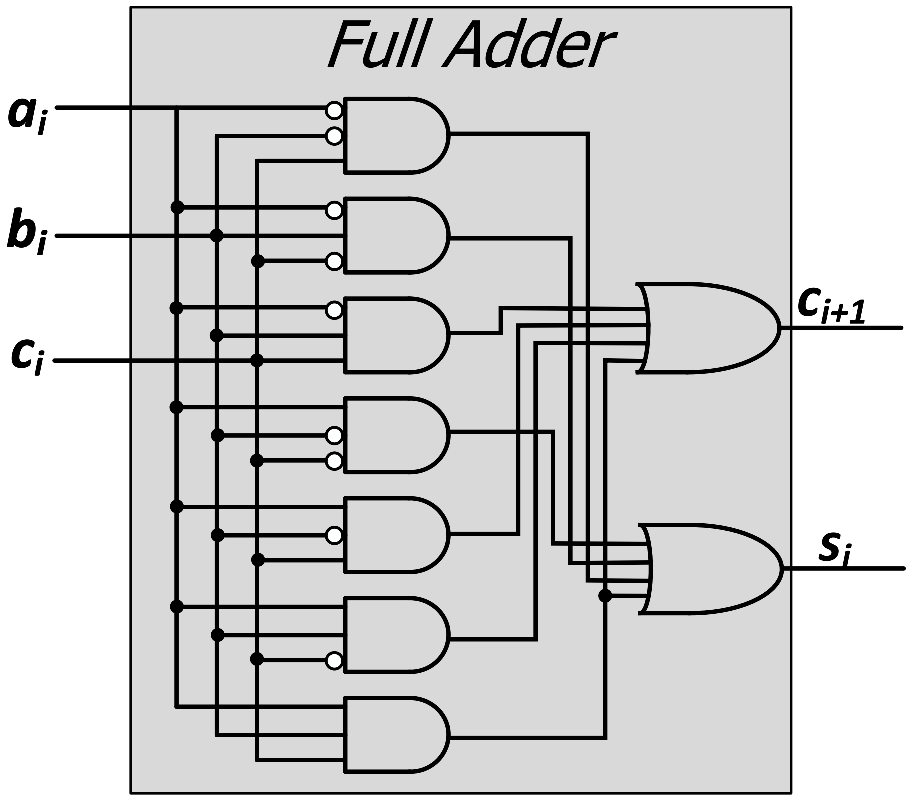


Figure 2.65 **Logic function using decoder**

Read H&H Chapter 2.8

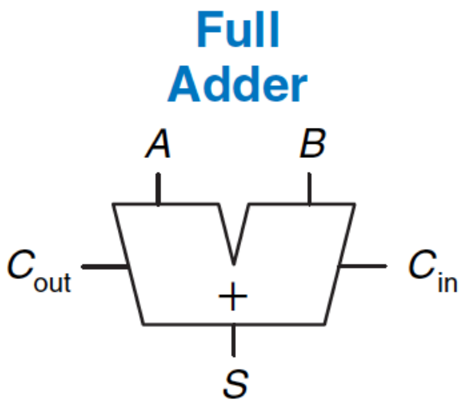
Logic Simplification using Boolean Algebra Rules

Recall: Full Adder in SOP Form Logic



a_i	b_i	$carry_i$	$carry_{i+1}$	S_i
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

Goal: Simplified Full Adder



C_{in}	A	B	C_{out}	S
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

$$S = A \oplus B \oplus C_{in}$$

$$C_{out} = AB + AC_{in} + BC_{in}$$

How do we simplify Boolean logic?

Quick Recap on Logic Simplification

- The original Boolean expression (i.e., logic circuit) may not be optimal

$$F = \sim A(A + B) + (B + AA)(A + \sim B)$$

- Can we reduce a given Boolean expression to an equivalent expression with fewer terms?

$$F = A + B$$

- The goal of logic simplification:
  - Reduce the number of gates/inputs
  - Reduce implementation cost

A basis for what the automated design tools are doing today

Logic Simplification

- Systematic techniques for simplifications
  - amenable to automation

Key Tool: The Uniting Theorem — $F = A\overline{B} + AB$

A	B	F
0	0	0
0		
1		
1	1	1

$$F = A\overline{B} + AB = A(\overline{B} + B) = A(1) = A$$

Essence of Simplification:

Find two element subsets of the ON-set where only one variable changes its value. This single varying variable *can be eliminated!*

value is not needed

→ B is eliminated, A remains

A	B	G
0	0	1
0	1	0
1	0	1
1	1	0

$$G = \overline{A}\,\overline{B} + A\overline{B} = (\overline{A} + A)\overline{B} = \overline{B}$$

B's value stays the same within the ON-set rows

A's value changes within the ON-set rows

→ A is eliminated, B remains

Logic Simplification: Karnaugh Maps (K-Maps)

Karnaugh Maps are Fun…

- A pictorial way of minimizing circuits by visualizing opportunities for simplification
- They are for you to **study on your own**…

- See Backup Slides
- Read H&H Section 2.7
- Watch videos of Lectures 5 and 6 from 2019 Digitech course:
  - https://youtu.be/0ks0PeaOUjE?list=PL5Q2soXY2Zi8J58xLKBNFQFHRO3GrXxA9&t=4570
  - https://youtu.be/ozs18ARNG6s?list=PL5Q2soXY2Zi8J58xLKBNFQFHRO3GrXxA9&t=220

Complex Cases

- One example

$$Cout = \bar{A}BC + A\bar{B}C + AB\bar{C} + ABC$$

- Problem
  - Easy to see how to apply Uniting Theorem…
  - Hard to know if you applied it in all the right places…
  - …especially in a function of many more variables

- Question
  - Is there an easier way to find potential simplifications?
  - i.e., potential applications of Uniting Theorem…?

- Answer
  - Need an intrinsically *geometric* representation for Boolean f( )
  - Something we can draw, see…

Karnaugh Map

- Karnaugh Map (K-map) method
  - K-map is an alternative method of representing the truth table that helps visualize adjacencies in up to 6 dimensions
  - Physical adjacency ↔ Logical adjacency

2-variable K-map

$A \backslash B$	0	1
0	00	01
1	10	11

3-variable K-map

$A \backslash BC$	00	01	11	10
0	000	001	011	010
1	100	101	111	110

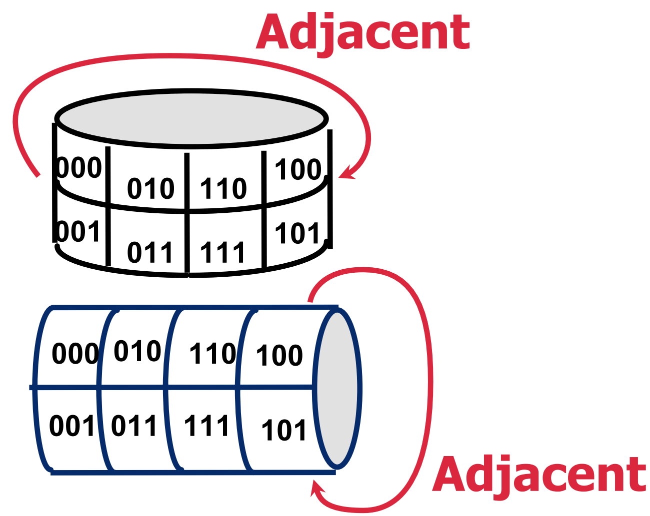
4-variable K-map

$AB \backslash CD$	00	01	11	10
00	0000	0001	0011	0010
01	0100	0101	0111	0110
11	1100	1101	1111	1110
10	1000	1001	1011	1010

Numbering Scheme: **00, 01, 11, 10 is called a "Gray Code" — only a *single bit (variable) changes* from one code word and the next code word**

Karnaugh Map Methods

A \ BC	00	01	11	10
0	000	001	011	010
1	100	101	111	110



K-map adjacencies go "around the edges"
Wrap around from first to last column
Wrap around from top row to bottom row

K-map Cover - 4 Input Variables

$AB \backslash CD$	00	01	11	10
00	1	0	0	1
01	0	1	0	0
11	1	1	1	1
10	1	1	1	1

$$\mathbf{F(A,B,C,D)} = \sum m(0,2,5,8,9,10,11,12,13,14,15)$$

$$\mathbf{F} = \mathbf{A} + \overline{B}\,\overline{D} + \mathbf{B}\overline{C}D$$

Strategy for "circling" rectangles on Kmap:

Biggest "oops!" that people forget:

Logic Minimization Using K-Maps

- Very simple guideline:
  - Circle all the rectangular blocks of 1's in the map, using the fewest possible number of circles
    - Each circle should be as large as possible
  - Read off the implicants that were circled

- More formally:
  - A Boolean equation is minimized when it is written as a sum of the fewest number of prime implicants
  - Each circle on the K-map represents an implicant
  - The largest possible circles are prime implicants

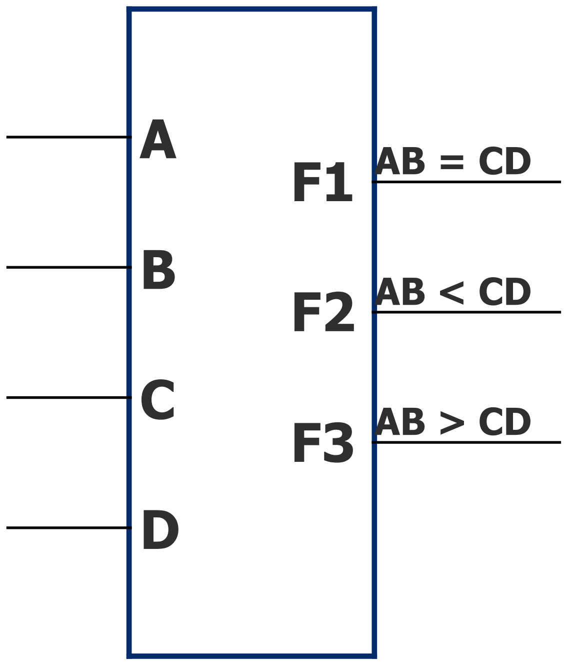
K-map Rules

- **What can be legally combined (circled) in the K-map?**
  - Rectangular groups of size 2^k for any integer k
  - Each cell has the same value (1, for now)
  - All values must be adjacent
    - Wrap-around edge is okay

- **How does a group become a term in an expression?**
  - Determine which literals are constant, and which vary across group
  - Eliminate varying literals, then AND the constant literals
    - constant 1 → use X, constant 0 → use $\overline{X}$

- **What is a good solution?**
  - Biggest groupings → eliminate more variables (literals) in each term
  - Fewest groupings → fewer terms (gates) all together
  - OR together all AND terms you create from individual groups

K-map Example: Two-bit Comparator



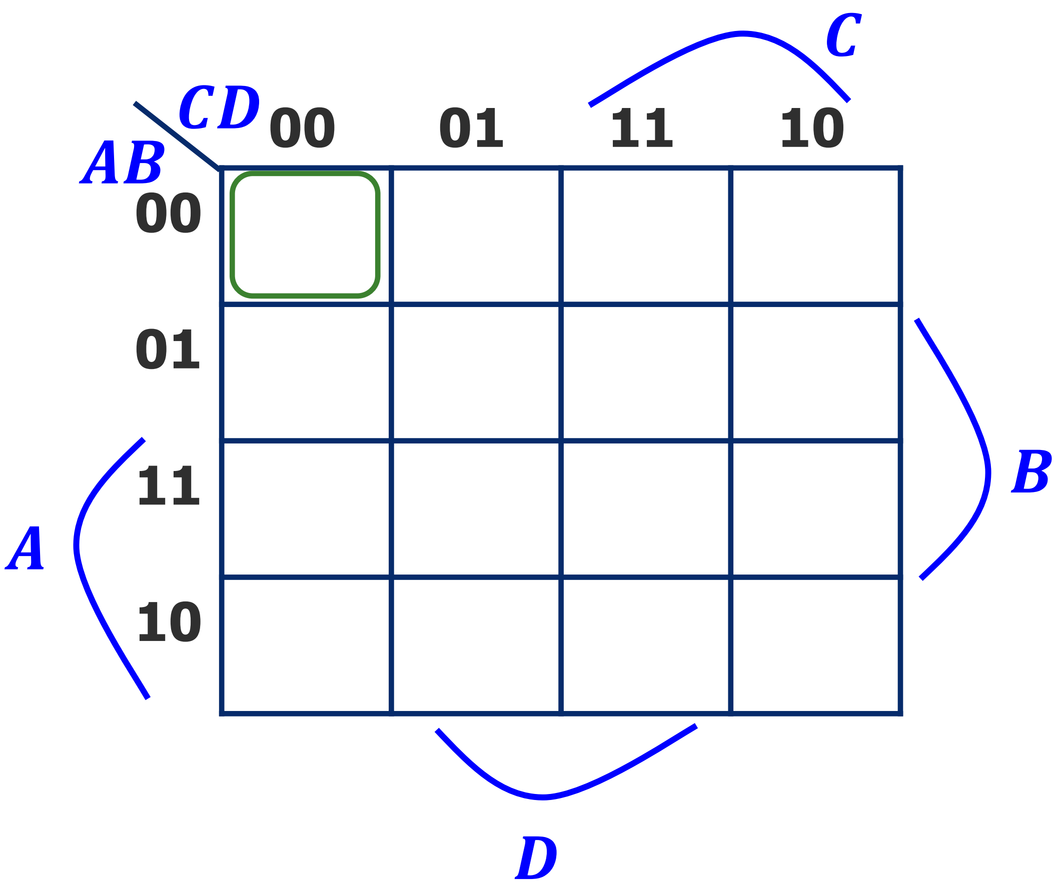
A	B	C	D	F1	F2	F3
0	0	0	0	1	0	0
0	0	0	1	0	1	0
0	0	1	0	0	1	0
0	0	1	1	0	1	0
0	1	0	0	0	0	1
0	1	0	1	1	0	0
0	1	1	0	0	1	0
0	1	1	1	0	1	0
1	0	0	0	0	0	1
1	0	0	1	0	0	1
1	0	1	0	1	0	0
1	0	1	1	0	1	0
1	1	0	0	0	0	1
1	1	0	1	0	0	1
1	1	1	0	0	0	1
1	1	1	1	1	0	0

Design Approach:

Write a 4-Variable K-map for each of the 3 output functions

K-map Example: Two-bit Comparator (2)

K-map for F1

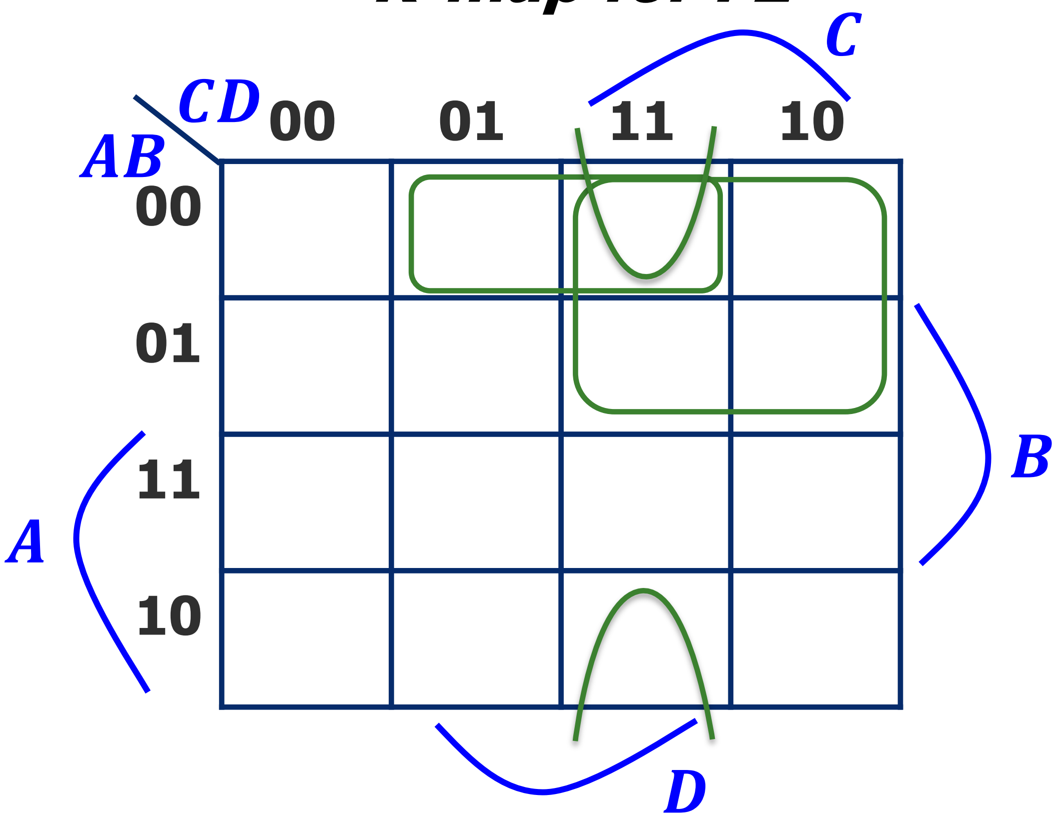


A	B	C	D	F1	F2	F3
0	0	0	0	1	0	0
0	0	0	1	0	1	0
0	0	1	0	0	1	0
0	0	1	1	0	1	0
0	1	0	0	0	0	1
0	1	0	1	1	0	0
0	1	1	0	0	1	0
0	1	1	1	0	1	0
1	0	0	0	0	0	1
1	0	0	1	0	0	1
1	0	1	0	1	0	0
1	0	1	1	0	1	0
1	1	0	0	0	0	1
1	1	0	1	0	0	1
1	1	1	0	0	0	1
1	1	1	1	1	0	0

F1 =

K-map Example: Two-bit Comparator (3)

K-map for F2



A	B	C	D	F1	F2	F3
0	0	0	0	1	0	0
0	0	0	1	0	1	0
0	0	1	0	0	1	0
0	0	1	1	0	1	0
0	1	0	0	0	0	1
0	1	0	1	1	0	0
0	1	1	0	0	1	0
0	1	1	1	0	1	0
1	0	0	0	0	0	1
1	0	0	1	0	0	1
1	0	1	0	1	0	0
1	0	1	1	0	1	0
1	1	0	0	0	0	1
1	1	0	1	0	0	1
1	1	1	0	0	0	1
1	1	1	1	1	0	0

F2 =

F3 = ? (Exercise for you)

K-maps with “Don’t Care”

- Don’t Care really means *I don’t care what my circuit outputs if this appears as input*
  - You have an engineering choice to use DON’T CARE patterns intelligently as 1 or 0 to better simplify the circuit

A	B	C	D	F	G
…					
0	1	1	0	X	X
0	1	1	1		
1	0	0	0	X	X
1	0	0	1		
…					

I can pick 00, 01, 10, 11 independently of below

I can pick 00, 01, 10, 11 independently of above

Example: BCD Increment Function

- BCD (Binary Coded Decimal) digits
  - Encode decimal digits 0 - 9 with bit patterns 0000_2 — 1001_2
  - When incremented, the decimal sequence is 0, 1, …, 8, 9, 0, 1

A	B	C	D	W	X	Y	Z
0	0	0	0	0	0	0	1
0	0	0	1	0	0	1	0
0	0	1	0	0	0	1	1
0	0	1	1	0	1	0	0
0	1	0	0	0	1	0	1
0	1	0	1	0	1	1	0
0	1	1	0	0	1	1	1
0	1	1	1	1	0	0	0
1	0	0	0	1	0	0	1
1	0	0	1	0	0	0	0
1	0	1	0	X	X	X	X
1	0	1	1	X	X	X	X
1	1	0	0	X	X	X	X
1	1	0	1	X	X	X	X
1	1	1	0	X	X	X	X
1	1	1	1	X	X	X	X

These input patterns should never be encountered in practice (hey -- it's a BCD number!) So, associated output values are "Don't Cares"

K-map for BCD Increment Function

A B

\+

W X

Z (without don’t cares) =

Z (with don’t cares) =

10	1		X	X

10			X	X

Y

AB \ CD	00	01	11	10
00		1		1
01		1		1
11	X	X	X	X
10			X	X

Z

AB \ CD	00	01	11	10
00	1			1
01	1			1
11	X	X	X	X
10	1		X	X

A B C D

K-map Summary

- Karnaugh maps as a formal systematic approach for logic simplification

- 2-, 3-, 4-variable K-maps

- K-maps with "Don't Care" outputs

- H&H Section 2.7

Digital Design & Computer Arch.

Lecture 4: Combinational Logic I

Prof. Onur Mutlu

ETH Zürich

Spring 2021

5 March 2021